



# Section I. MAX II Device Family Data Sheet

This section provides designers with the data sheet specifications for MAX<sup>®</sup> II devices. The chapters contain feature definitions of the internal architecture, Joint Test Action Group (JTAG) and in-system programmability (ISP) information, DC operating conditions, AC timing parameters, and ordering information for MAX II devices.

This section includes the following chapters:

- [Chapter 1. Introduction](#)
- [Chapter 2. MAX II Architecture](#)
- [Chapter 3. JTAG & In-System Programmability](#)
- [Chapter 4. Hot Socketing & Power-On Reset in MAX II Devices](#)
- [Chapter 5. DC & Switching Characteristics](#)
- [Chapter 6. Reference & Ordering Information](#)

## Revision History

The table below shows the revision history for [Chapters 1](#) through [6](#).

Chapter(s)	Date/Version	Changes Made
1	June 2004, v1.1	Updated timing numbers in Table 1.
2	June 2004, v1.1	Added CFM acronym. Corrected Figure 2-19.
3	June 2004, v1.1	Corrected Figure 3-1. Added CFM acronym.
4	June 2004, v1.1	Corrected Figure 4-2.
5	June 2004, v1.1	Updated timing tables 5-15 through 5-32.
6	March 2004, v1.0	Initial Release.



## Introduction

The MAX<sup>®</sup> II family of instant-on, non-volatile CPLDs is based on a 0.18- $\mu$ m, 6-layer-metal-flash process, with densities from 240 to 2,210 logic elements (LEs) (128 to 2,210 equivalent macrocells) and non-volatile storage of 8 Kbits. MAX II devices offer high I/O counts, fast performance, and reliable fitting versus other CPLD architectures. Featuring MultiVolt<sup>™</sup> core, a user flash memory (UFM) block, and enhanced in-system programmability (ISP), MAX II devices are designed to reduce cost and power while providing programmable solutions for applications such as bus bridging, I/O expansion, power-on reset (POR) and sequencing control, and device configuration control.

The following shows the main sections of the MAX II CPLD Family Data Sheet:

Section	Page
Features . . . . .	1-2
Functional Description . . . . .	2-1
Logic Array Blocks . . . . .	2-4
Logic Elements . . . . .	2-7
MultiTrack Interconnect . . . . .	2-15
Global Signals . . . . .	2-20
User Flash Memory Block . . . . .	2-23
MultiVolt Core . . . . .	2-27
I/O Structure . . . . .	2-28
IEEE Std. 1149.1 (JTAG) Boundary Scan Support . . . . .	3-1
In System Programmability . . . . .	3-4
Hot Socketing . . . . .	4-1
Power-On Reset Circuitry . . . . .	4-6
Operating Conditions . . . . .	5-1
Power Consumption . . . . .	5-8
Timing Model & Specifications . . . . .	5-8
Device Pin-Outs . . . . .	6-1
Ordering Information . . . . .	6-1

## Features

- Low-cost, low-power CPLD
- Instant-on, non-volatile architecture
- Standby current as low as 2 mA
- Provides fast propagation delay and clock-to-output times
- Provides four global clocks with two clocks available per logic array block (LAB)
- UFM block up to 8 Kbits for non-volatile storage
- MultiVolt core enabling external supply voltages to the device of either 3.3 V/2.5 V or 1.8 V
- MultiVolt I/O interface supporting 3.3-V, 2.5-V, 1.8-V, and 1.5-V logic levels
- Bus-friendly architecture including programmable slew rate, drive strength, bus-hold, and programmable pull-up resistors
- Schmitt triggers enabling noise tolerant inputs (programmable per pin)
- Fully compliant with the Peripheral Component Interconnect Special Interest Group (PCI SIG) *PCI Local Bus Specification, Revision 2.2* for 3.3-V operation at 33 MHz
- Supports hot-socketing
- Built-in Joint Test Action Group (JTAG) boundary-scan test (BST) circuitry compliant with IEEE Std. 1149.1-1990
- ISP circuitry compliant with IEEE Std. 1532

Table 1–1 shows MAX II device features.

Feature	EPM240	EPM570	EPM1270	EPM2210
LEs	240	570	1,270	2,210
Typical Equivalent Macrocells	192	440	980	1,700
Equivalent Macrocell Range	128 to 240	240 to 570	570 to 1,270	1,270 to 2,210
UFM Size (bits)	8,192	8,192	8,192	8,192
Maximum User I/O pins	80	160	212	272
$t_{PD1}$ (ns) (1)	4.4	5.2	6.1	6.9
$f_{CNT}$ (MHz) (2)	304	304	304	304
$t_{SU}$ (ns)	1.9	1.8	1.7	1.6
$t_{CO}$ (ns)	4.2	4.3	4.4	4.5

### Notes to Table 1–1:

- (1)  $t_{PD1}$  represents a pin-to-pin delay for the worst case I/O placement with a full diagonal path across the device and combinational logic implemented in a single LUT and LAB that is adjacent to the output pin.
- (2) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay will run faster than this number.



For more information on equivalent macrocells, refer to the *MAX II Logic Element to Macrocell Conversion Methodology* white paper.

MAX II devices are available in three speed grades: -3, -4, -5 with -3 being the fastest. These speed grades represent overall relative performance, not any specific timing parameter. For propagation delay timing numbers within each speed grade and density, see the chapter on *DC & Switching Characteristics*. [Table 1-2](#) shows MAX II device speed-grade offerings.

Device	Speed Grade		
	-3	-4	-5
EPM240	✓	✓	✓
EPM570	✓	✓	✓
EPM1270	✓	✓	✓
EPM2210	✓	✓	✓

MAX II devices are available in space-saving FineLine BGA® and thin quad flat pack (TQFP) packages (see [Tables 1-3](#) and [1-4](#)). MAX II devices support vertical migration within the same package (e.g., you can migrate between the EPM570, EPM1270, and EPM2210 devices in the 256-pin FineLine BGA package). Vertical migration means that you can migrate to devices whose dedicated pins and JTAG pins are the same and power pins are subsets or supersets for a given package across device densities. The largest density in any package has the highest number of power pins; you must layout for the largest planned density in a package to provide the necessary power pins for migration. For I/O pin migration across densities, cross reference the available I/O pins using the device pin-outs for all planned densities of a given package type to identify which I/O pins can be migrated. The Quartus® II software can automatically cross reference and place all pins for you when given a device migration list.

Device	100-Pin TQFP	144-Pin TQFP	256-Pin FineLine BGA	324-Pin FineLine BGA
EPM240	80			
EPM570	76	116	160	
EPM1270		116	212	
EPM2210			204	272

**Table 1–4. MAX II TQFP & FineLine BGA Package Sizes**

Package	100-Pin TQFP	144-Pin TQFP	256-Pin FineLine BGA	324-Pin FineLine BGA
Pitch (mm)	0.5	0.5	1	1
Area (mm <sup>2</sup> )	256	484	289	361
Length x width (mm x mm)	16 x 16	22 x 22	17 x 17	19 x 19

MAX II devices have an internal linear voltage regulator which supports external supply voltages of 3.3 V or 2.5 V, regulating the supply down to the internal operating voltage of 1.8 V. MAX II devices with a “G” in the ordering code only accept 1.8 V as external supply voltage. [Table 1–5](#) shows the external supply voltages supported by the MAX II family.

**Table 1–5. MAX II External Supply Voltages**

Devices	EPM240 EPM570 EPM1270 EPM2210	EPM240G EPM570G EPM1270G EPM2210G (1)
MultiVolt core external supply voltage ( $V_{CCINT}$ ) (2)	3.3 V, 2.5 V	1.8 V
MultiVolt I/O interface voltage levels ( $V_{CCIO}$ )	1.5 V, 1.8 V, 2.5 V, 3.3 V	1.5 V, 1.8 V, 2.5 V, 3.3 V

**Notes to Table 1–5:**

- (1) MAX II devices with a “G” in the ordering code do not have an internal voltage regulator and only accept 1.8 V on their  $V_{CCINT}$  pins. Contact Altera for availability on these devices.
- (2) MAX II devices operate internally at 1.8 V.

### Functional Description

MAX® II devices contain a two-dimensional row- and column-based architecture to implement custom logic. Column and row interconnect provide signal interconnects between the logic array blocks (LABs).

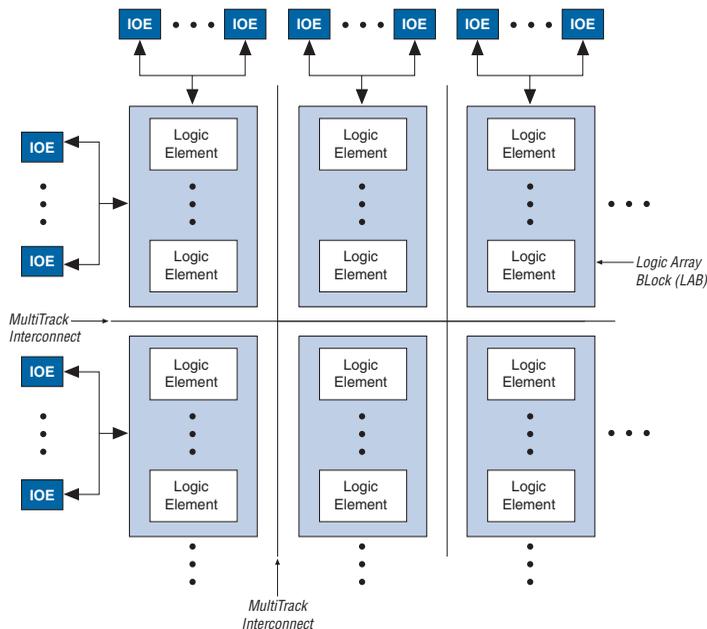
The logic array consists of LABs, with 10 logic elements (LEs) in each LAB. An LE is a small unit of logic providing efficient implementation of user logic functions. LABs are grouped into rows and columns across the device. The MultiTrack™ interconnect provides fast granular timing delays between LABs. The fast routing between LEs provides minimum timing delay for added levels of logic versus globally routed interconnect structures.

The MAX II device I/O pins are fed by an I/O element (IOE) located at the ends of LAB rows and columns around the periphery of the device. Each IOE contains a bidirectional I/O buffer with several advanced features. I/O pins support Schmitt trigger inputs and various single-ended standards, such as 33-MHz, 32-bit PCI and LVTTTL.

MAX II devices provide a global clock network. The global clock network consists of four global clock lines that drive throughout the entire device, providing clocks for all resources within the device. The global clock lines can also be used for control signals such as clear, preset, or output enable.

Figure 2-1 shows a functional block diagram of the MAX II device.

**Figure 2–1. MAX II Device Block Diagram**



Each MAX II device contains a flash memory block within its floorplan. On the EPM240 device, this block is located on the left side of the device. For the EPM570, EPM1270, and EPM2210 devices, the flash memory block is located on the bottom-left area of the device. The majority of this flash memory storage is partitioned as the dedicated configuration flash memory (CFM) block. The CFM block provides the non-volatile storage for all of the SRAM configuration information. The CFM automatically downloads and configures the logic and I/O at power-up providing instant-on operation.



See *Hot Socketing & Power-On Reset in MAX II Devices* for more information on configuration upon power-up.

A portion of the flash memory within the MAX II device is partitioned into a small block for user data. This user flash memory (UFM) block provides 8,192 bits of general-purpose user storage. The UFM provides programmable port connections to the logic array for reading and for writing. There are three LAB rows adjacent to this block, with column numbers varying by device.

Table 2-1 shows the number of LAB rows and columns in each device as well as the number of LAB rows and columns adjacent to the flash memory area in the EPM570, EPM1270, and EPM2210 devices. The long LAB rows are full LAB rows that extend from one side of row I/O blocks to the other. The short LAB rows are adjacent to the UFM block; their length is shown as width in LAB columns.

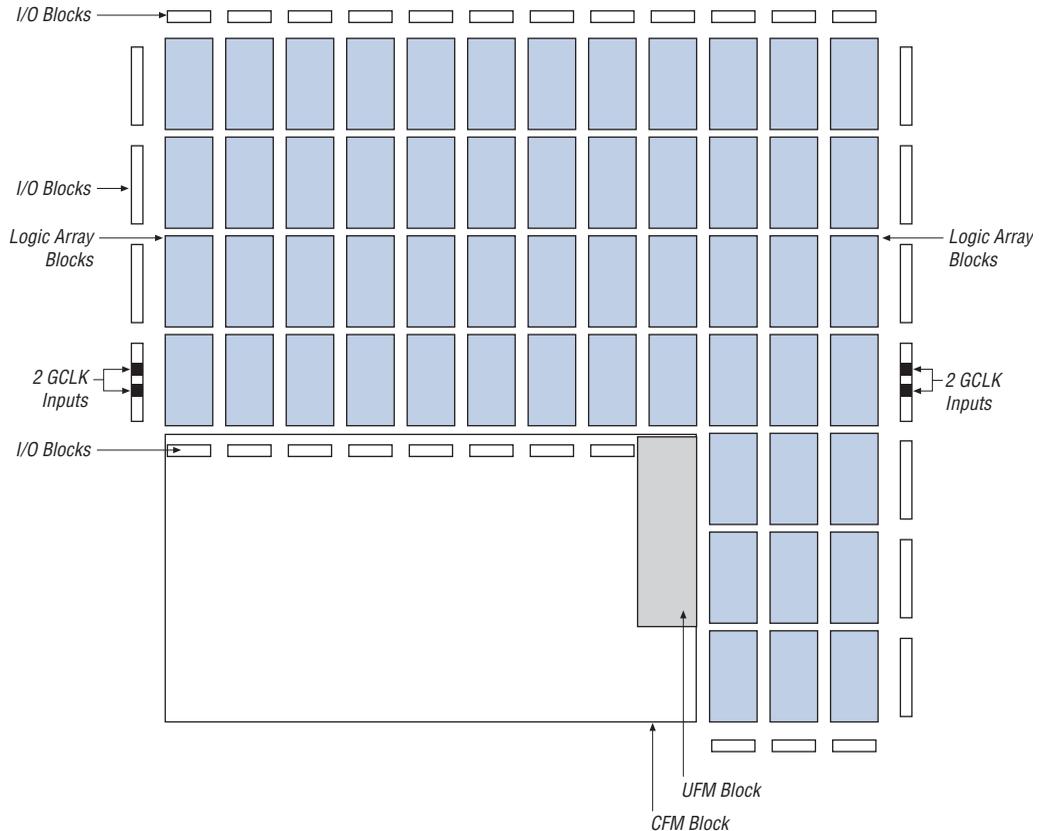
Devices	UFM Blocks	LAB Columns	LAB Rows		Total LABs
			Long LAB Rows	Short LAB Rows (Width) (1)	
EPM240	1	6	4	-	24
EPM570	1	12	4	3 (3)	57
EPM1270	1	16	7	3 (5)	127
EPM2210	1	20	10	3 (7)	221

**Note to Table 2-1:**

(1) The width is the number of LAB columns in length.

Figure 2-2 shows a floorplan of a MAX II device.

**Figure 2–2. MAX II Device Floorplan** *Note (1)*



**Note to Figure 2–2:**

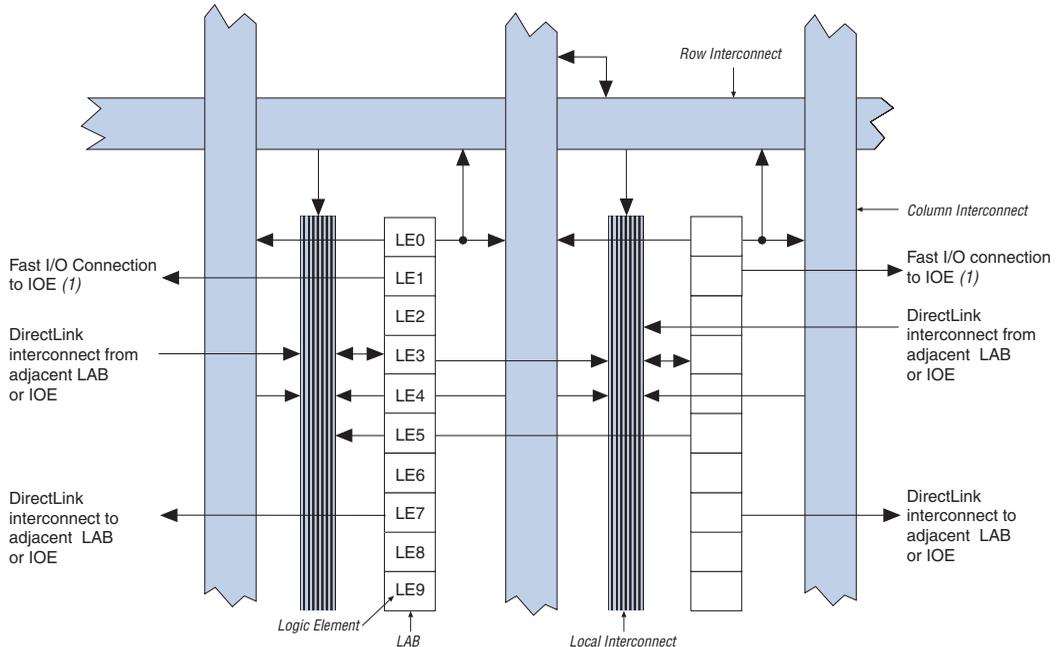
- (1) The device shown is an EPM570 device. EPM1270 and EPM2210 devices have a similar floorplan with more LABs. For the EPM240 devices, the CFM and UFM block is rotated left 90 degrees covering the left side of the device.

## Logic Array Blocks

Each LAB consists of 10 LEs, LE carry chains, LAB control signals, a local interconnect, a look-up table (LUT) chain, and register chain connection lines. There are 26 possible unique inputs into an LAB, with an additional 10 local feedback input lines fed by LE outputs in the same LAB. The local interconnect transfers signals between LEs in the same LAB. LUT chain connections transfer the output of one LE's LUT to the adjacent LE for fast sequential LUT connections within the same LAB. Register chain connections transfer the output of one LE's register to the adjacent LE's register within an LAB. The Quartus® II software places associated logic

within an LAB or adjacent LABs, allowing the use of local, LUT chain, and register chain connections for performance and area efficiency. Figure 2-3 shows the MAX II LAB.

**Figure 2-3. MAX II LAB Structure**



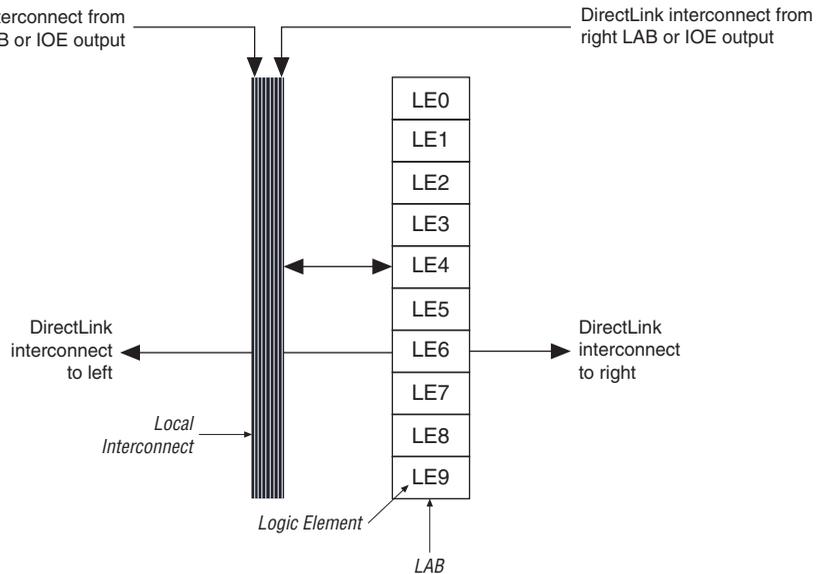
**Note to Figure 2-3:**

(1) Only from LABs adjacent to IOEs.

## LAB Interconnects

The LAB local interconnect can drive LEs within the same LAB. The LAB local interconnect is driven by column and row interconnects and LE outputs within the same LAB. Neighboring LABs, from the left and right can also drive an LAB's local interconnect through the DirectLink connection. The DirectLink connection feature minimizes the use of row and column interconnects, providing higher performance and flexibility. Each LE can drive 30 other LEs through fast local and DirectLink interconnects. Figure 2-4 shows the DirectLink connection.

**Figure 2–4. DirectLink Connection**



## LAB Control Signals

Each LAB contains dedicated logic for driving control signals to its LEs. The control signals include two clocks, two clock enables, two asynchronous clears, a synchronous clear, an asynchronous preset/load, a synchronous load, and add/subtract control signals, providing a maximum of 10 control signals at a time. Although synchronous load and clear signals are generally used when implementing counters, they can also be used with other functions.

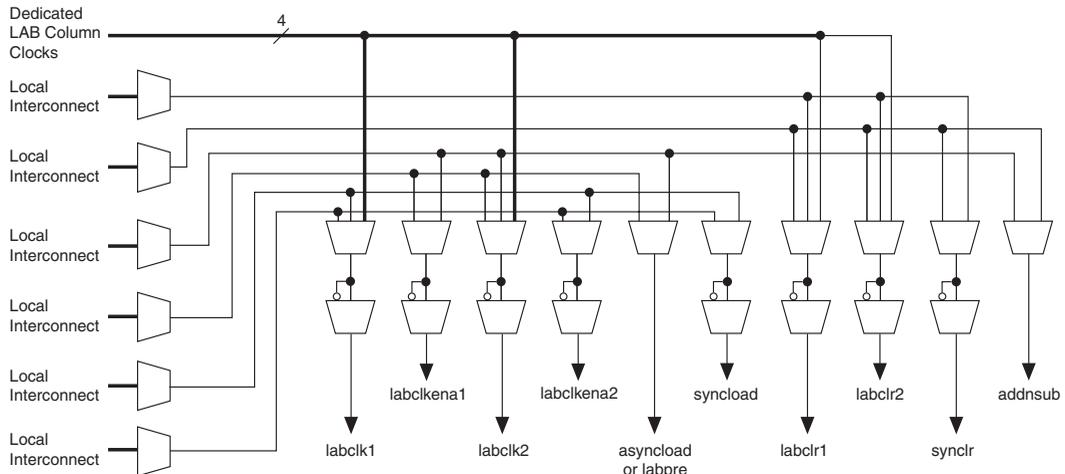
Each LAB can use two clocks and two clock enable signals. Each LAB's clock and clock enable signals are linked. For example, any LE in a particular LAB using the `labclk1` signal also uses `labckena1`. If the LAB uses both the rising and falling edges of a clock, it also uses both LAB-wide clock signals. De-asserting the clock enable signal turns off the LAB-wide clock.

Each LAB can use two asynchronous clear signals and an asynchronous load/preset signal. The asynchronous load acts as a preset when the asynchronous load data input is tied high.

With the LAB-wide `addnsub` control signal, a single LE can implement a one-bit adder and subtractor. This saves LE resources and improves performance for logic functions such as correlators and signed multipliers that alternate between addition and subtraction depending on data.

The LAB column clocks [3..0], driven by the global clock network, and LAB local interconnect generate the LAB-wide control signals. The MultiTrack™ interconnect structure drives the LAB local interconnect for non-global control signal generation. The MultiTrack interconnect's inherent low skew allows clock and control signal distribution in addition to data. [Figure 2-5](#) shows the LAB control signal generation circuit.

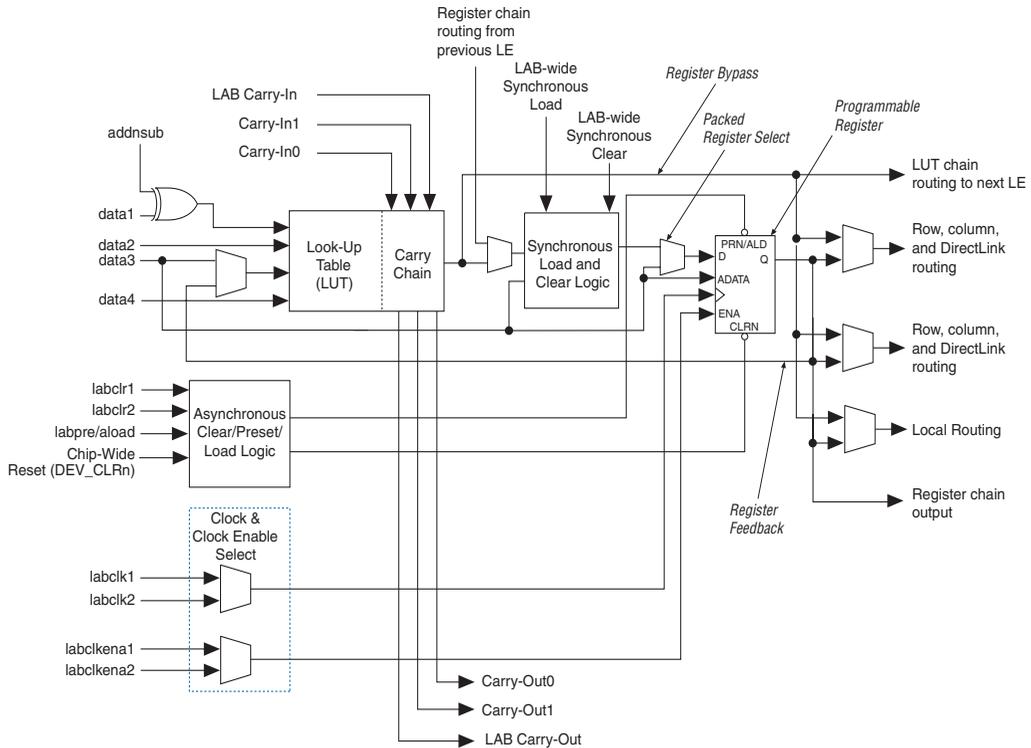
**Figure 2-5. LAB-Wide Control Signals**



## Logic Elements

The smallest unit of logic in the MAX II architecture, the LE, is compact and provides advanced features with efficient logic utilization. Each LE contains a four-input LUT, which is a function generator that can implement any function of four variables. In addition, each LE contains a programmable register with carry chain with carry select capability. A single LE also supports dynamic single bit addition or subtraction mode selectable by an LAB-wide control signal. Each LE drives all types of interconnects: local, row, column, LUT chain, register chain, and DirectLink interconnects. See [Figure 2-6](#).

Figure 2–6. MAX II LE



Each LE’s programmable register can be configured for D, T, JK, or SR operation. Each register has data, true asynchronous load data, clock, clock enable, clear, and asynchronous load/preset inputs. Global signals, general-purpose I/O pins, or any LE can drive the register’s clock and clear control signals. Either general-purpose I/O pins or LEs can drive the clock enable, preset, asynchronous load, and asynchronous data. The asynchronous load data input comes from the `data3` input of the LE. For combinational functions, the LUT output bypasses the register and drives directly to the LE outputs.

Each LE has three outputs that drive the local, row, and column routing resources. The LUT or register output can drive these three outputs independently. Two LE outputs drive column or row and DirectLink routing connections and one drives local interconnect resources. This allows the LUT to drive one output while the register drives another output. This register packing feature improves device utilization because the device can use the register and the LUT for unrelated functions. Another special packing mode allows the register output to feed back into

the LUT of the same LE so that the register is packed with its own fan-out LUT. This provides another mechanism for improved fitting. The LE can also drive out registered and unregistered versions of the LUT output.

## LUT Chain & Register Chain

In addition to the three general routing outputs, the LEs within an LAB have LUT chain and register chain outputs. LUT chain connections allow LUTs within the same LAB to cascade together for wide input functions. Register chain outputs allow registers within the same LAB to cascade together. The register chain output allows an LAB to use LUTs for a single combinational function and the registers to be used for an unrelated shift register implementation. These resources speed up connections between LABs while saving local interconnect resources. See “[MultiTrack Interconnect](#)” on page 2–15 for more information on LUT chain and register chain connections.

## addsub Signal

The LE’s dynamic adder/subtractor feature saves logic resources by using one set of LEs to implement both an adder and a subtractor. This feature is controlled by the LAB-wide control signal `addsub`. The `addsub` signal sets the LAB to perform either  $A + B$  or  $A - B$ . The LUT computes addition; subtraction is computed by adding the two’s complement of the intended subtractor. The LAB-wide signal converts to two’s complement by inverting the B bits within the LAB and setting carry-in to 1, which adds one to the least significant bit (LSB). The LSB of an adder/subtractor must be placed in the first LE of the LAB, where the LAB-wide `addsub` signal automatically sets the carry-in to 1. The Quartus II Compiler automatically places and uses the adder/subtractor feature when using adder/subtractor parameterized functions.

## LE Operating Modes

The MAX II LE can operate in one of the following modes:

- Normal mode
- Dynamic arithmetic mode

Each mode uses LE resources differently. In each mode, eight available inputs to the LE, the four data inputs from the LAB local interconnect, `carry-in0` and `carry-in1` from the previous LE, the LAB carry-in from the previous carry-chain LAB, and the register chain connection are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, asynchronous

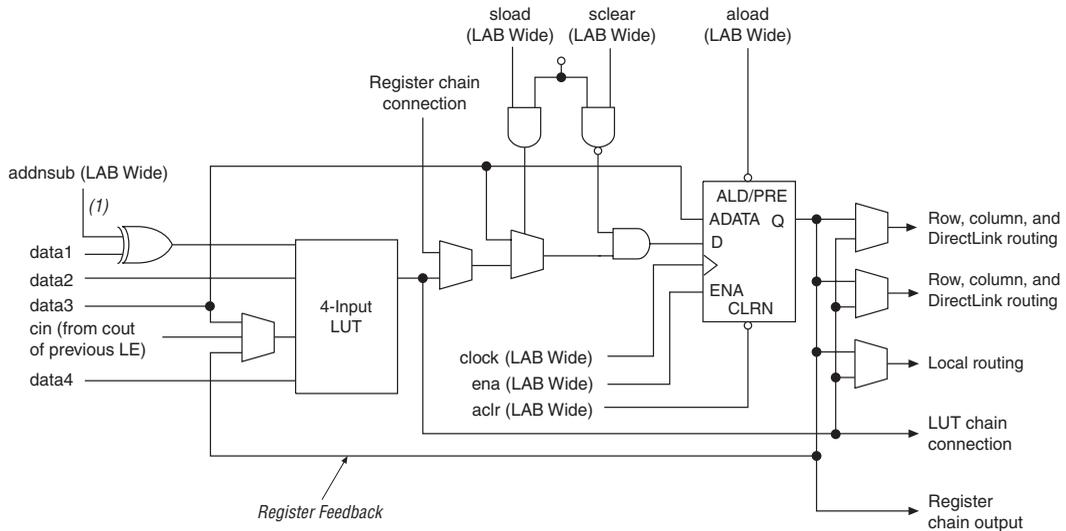
preset/load, synchronous clear, synchronous load, and clock enable control for the register. These LAB-wide signals are available in all LE modes. The addnsub control signal is allowed in arithmetic mode.

The Quartus II software, in conjunction with parameterized functions such as library of parameterized modules (LPM) functions, automatically chooses the appropriate mode for common functions such as counters, adders, subtractors, and arithmetic functions.

*Normal Mode*

The normal mode is suitable for general logic applications and combinational functions. In normal mode, four data inputs from the LAB local interconnect are inputs to a four-input LUT (see Figure 2-7). The Quartus II Compiler automatically selects the carry-in or the data3 signal as one of the inputs to the LUT. Each LE can use LUT chain connections to drive its combinational output directly to the next LE in the LAB. Asynchronous load data for the register comes from the data3 input of the LE. LEs in normal mode support packed registers.

**Figure 2-7. LE in Normal Mode**



**Note to Figure 2-7:**

(1) This signal is only allowed in normal mode if the LE is at the end of an adder/subtractor chain.

### *Dynamic Arithmetic Mode*

The dynamic arithmetic mode is ideal for implementing adders, counters, accumulators, wide parity functions, and comparators. An LE in dynamic arithmetic mode uses four 2-input LUTs configurable as a dynamic adder/subtractor. The first two 2-input LUTs compute two summations based on a possible carry-in of 1 or 0; the other two LUTs generate carry outputs for the two chains of the carry select circuitry. As shown in [Figure 2-8](#), the LAB carry-in signal selects either the `carry-in0` or `carry-in1` chain. The selected chain's logic level in turn determines which parallel sum is generated as a combinational or registered output. For example, when implementing an adder, the sum output is the selection of two possible calculated sums:

```
data1 + data2 + carry in0
```

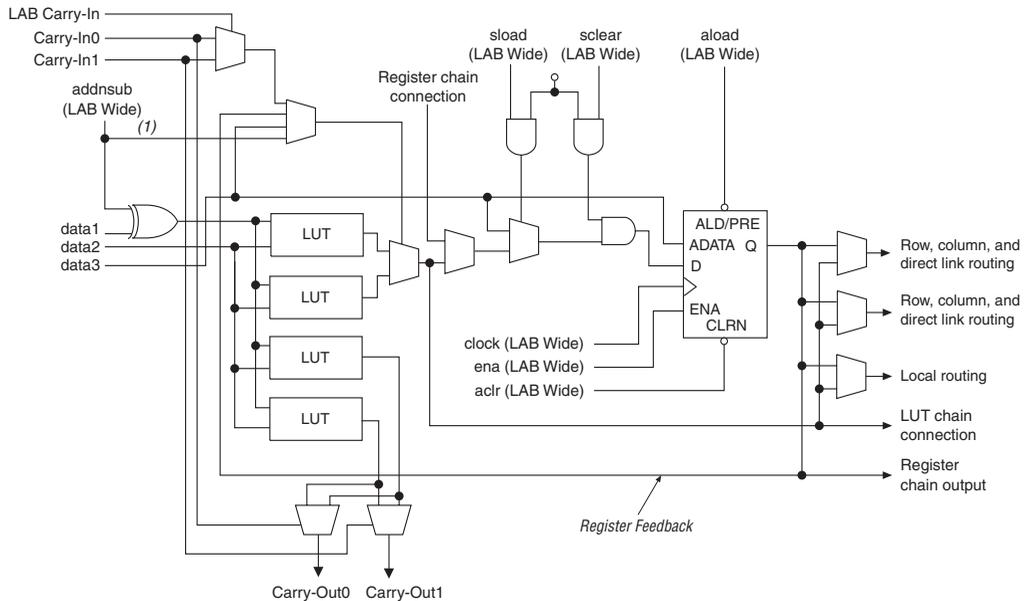
or

```
data1 + data2 + carry-in1
```

The other two LUTs use the `data1` and `data2` signals to generate two possible carry-out signals: one for a carry of 1 and the other for a carry of 0. The `carry-in0` signal acts as the carry select for the `carry-out0` output and `carry-in1` acts as the carry select for the `carry-out1` output. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

The dynamic arithmetic mode also offers clock enable, counter enable, synchronous up/down control, synchronous clear, synchronous load, and dynamic adder/subtractor options. The LAB local interconnect data inputs generate the counter enable and synchronous up/down control signals. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. The Quartus II software automatically places any registers that are not used by the counter into other LABs. The `addnsub` LAB-wide signal controls whether the LE acts as an adder or subtractor.

**Figure 2–8. LE in Dynamic Arithmetic Mode**



**Note to Figure 2–8:**

(1) The addnsub signal is tied to the carry input for the first LE of a carry chain only.

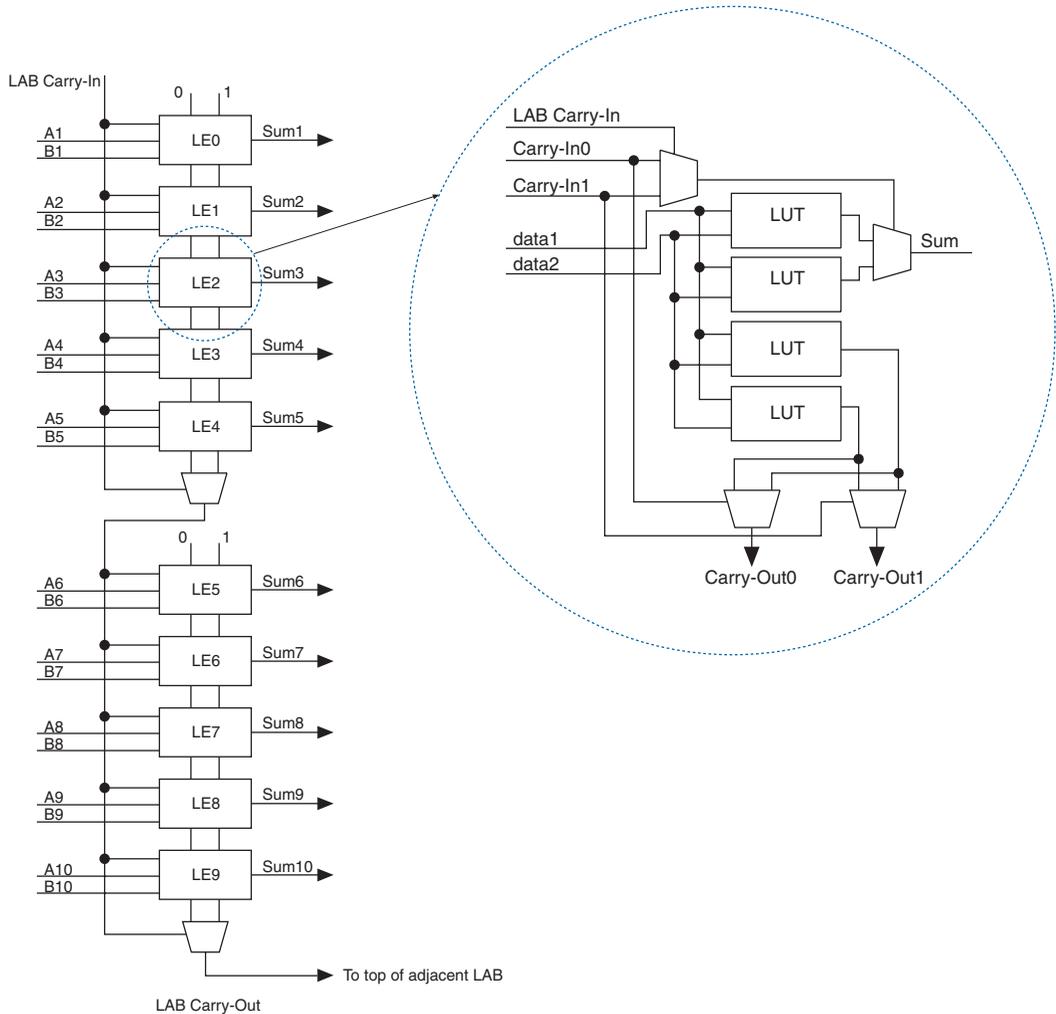
### Carry-Select Chain

The carry-select chain provides a very fast carry-select function between LEs in dynamic arithmetic mode. The carry-select chain uses the redundant carry calculation to increase the speed of carry functions. The LE is configured to calculate outputs for a possible carry-in of 0 and carry-in of 1 in parallel. The carry-in0 and carry-in1 signals from a lower-order bit feed forward into the higher-order bit via the parallel carry chain and feed into both the LUT and the next portion of the carry chain. Carry-select chains can begin in any LE within an LAB.

The speed advantage of the carry-select chain is in the parallel pre-computation of carry chains. Since the LAB carry-in selects the precomputed carry chain, not every LE is in the critical path. Only the propagation delays between LAB carry-in generation (LE 5 and LE 10) are now part of the critical path. This feature allows the MAX II architecture to implement high-speed counters, adders, multipliers, parity functions, and comparators of arbitrary width.

Figure 2-9 shows the carry-select circuitry in an LAB for a 10-bit full adder. One portion of the LUT generates the sum of two bits using the input signals and the appropriate carry-in bit; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for accumulator functions. Another portion of the LUT generates carry-out bits. An LAB-wide carry-in bit selects which chain is used for the addition of given inputs. The carry-in signal for each chain, `carry-in0` or `carry-in1`, selects the carry-out to carry forward to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it is fed to local, row, or column interconnects.

**Figure 2–9. Carry Select Chain**



The Quartus II software automatically creates carry chain logic during design processing, or the designer can create it manually during design entry. Parameterized functions such as LPM functions automatically take advantage of carry chains for the appropriate functions. The Quartus II software creates carry chains longer than 10 LEs by linking adjacent LABs within the same row together automatically. A carry chain can extend horizontally up to one full LAB row, but they do not extend between LAB rows.

### *Clear & Preset Logic Control*

LAB-wide signals control the logic for the register's clear and preset signals. The LE directly supports an asynchronous clear and preset function. The register preset is achieved through the asynchronous load of a logic high. The direct asynchronous preset does not require a NOT-gate push-back technique. MAX II devices support simultaneous preset/asynchronous load and clear signals. An asynchronous clear signal takes precedence if both signals are asserted simultaneously. Each LAB supports up to two clears and one preset signal.

In addition to the clear and preset ports, MAX II devices provide a chip-wide reset pin (`DEV_CLRn`) that resets all registers in the device. An option set before compilation in the Quartus II software controls this pin. This chip-wide reset overrides all other control signals. Driving this signal low before or during power-up prevents user mode from releasing clears within the design. This allows you to control when clear is released on a device that has just been powered-up. If not set for its chip-wide reset function, the `DEV_CLRn` pin is a regular I/O pin.

## **MultiTrack Interconnect**

In the MAX II architecture, connections between LEs, the UFM, and device I/O pins are provided by the MultiTrack interconnect structure. The MultiTrack interconnect consists of continuous, performance-optimized routing lines used for inter- and intra-design block connectivity. The Quartus II Compiler automatically places critical design paths on faster interconnects to improve design performance.

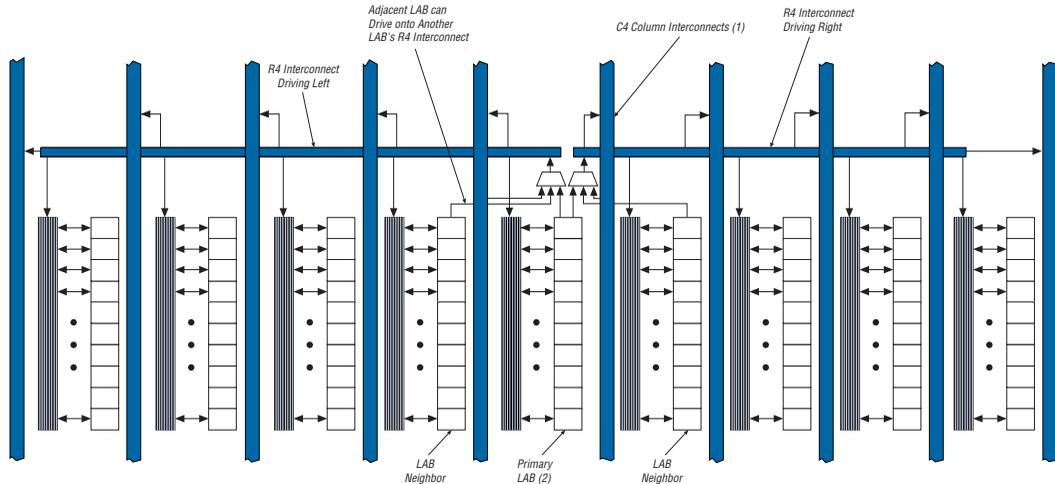
The MultiTrack interconnect consists of row and column interconnects that span fixed distances. A routing structure with fixed length resources for all devices allows predictable and short delays between logic levels instead of large delays associated with global or long routing lines. Dedicated row interconnects route signals to and from LABs within the same row. These row resources include:

- DirectLink interconnects between LABs
- R4 interconnects traversing four LABs to the right or left

The DirectLink interconnect allows an LAB to drive into the local interconnect of its left and right neighbors. The DirectLink interconnect provides fast communication between adjacent LABs and/or blocks without using row interconnect resources.

The R4 interconnects span four LABs and are used for fast row connections in a four-LAB region. Every LAB has its own set of R4 interconnects to drive either left or right. Figure 2–10 shows R4 interconnect connections from an LAB. R4 interconnects can drive and be driven by row IOEs. For LAB interfacing, a primary LAB or horizontal LAB neighbor can drive a given R4 interconnect. For R4 interconnects that drive to the right, the primary LAB and right neighbor can drive on to the interconnect. For R4 interconnects that drive to the left, the primary LAB and its left neighbor can drive on to the interconnect. R4 interconnects can drive other R4 interconnects to extend the range of LABs they can drive. R4 interconnects can also drive C4 interconnects for connections from one row to another.

Figure 2–10. R4 Interconnect Connections



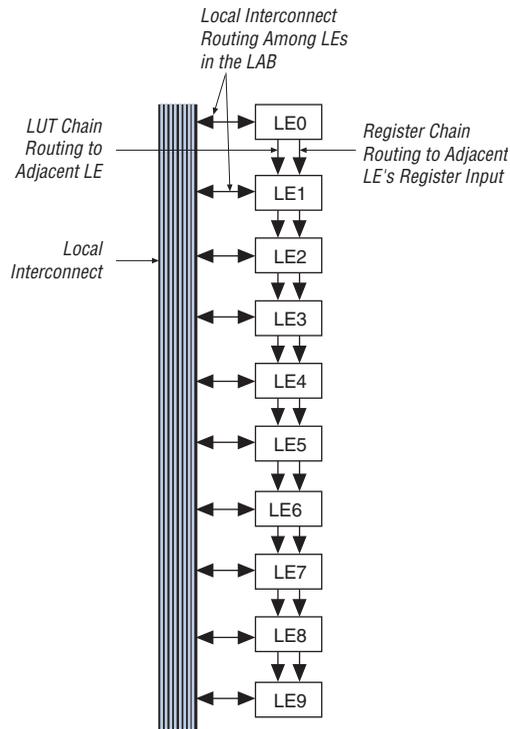
Notes to Figure 2–10:

- (1) C4 interconnects can drive R4 interconnects.
- (2) This pattern is repeated for every LAB in the LAB row.

The column interconnect operates similarly to the row interconnect. Each column of LABs is served by a dedicated column interconnect, which vertically routes signals to and from LABs and row and column IOEs. These column resources include:

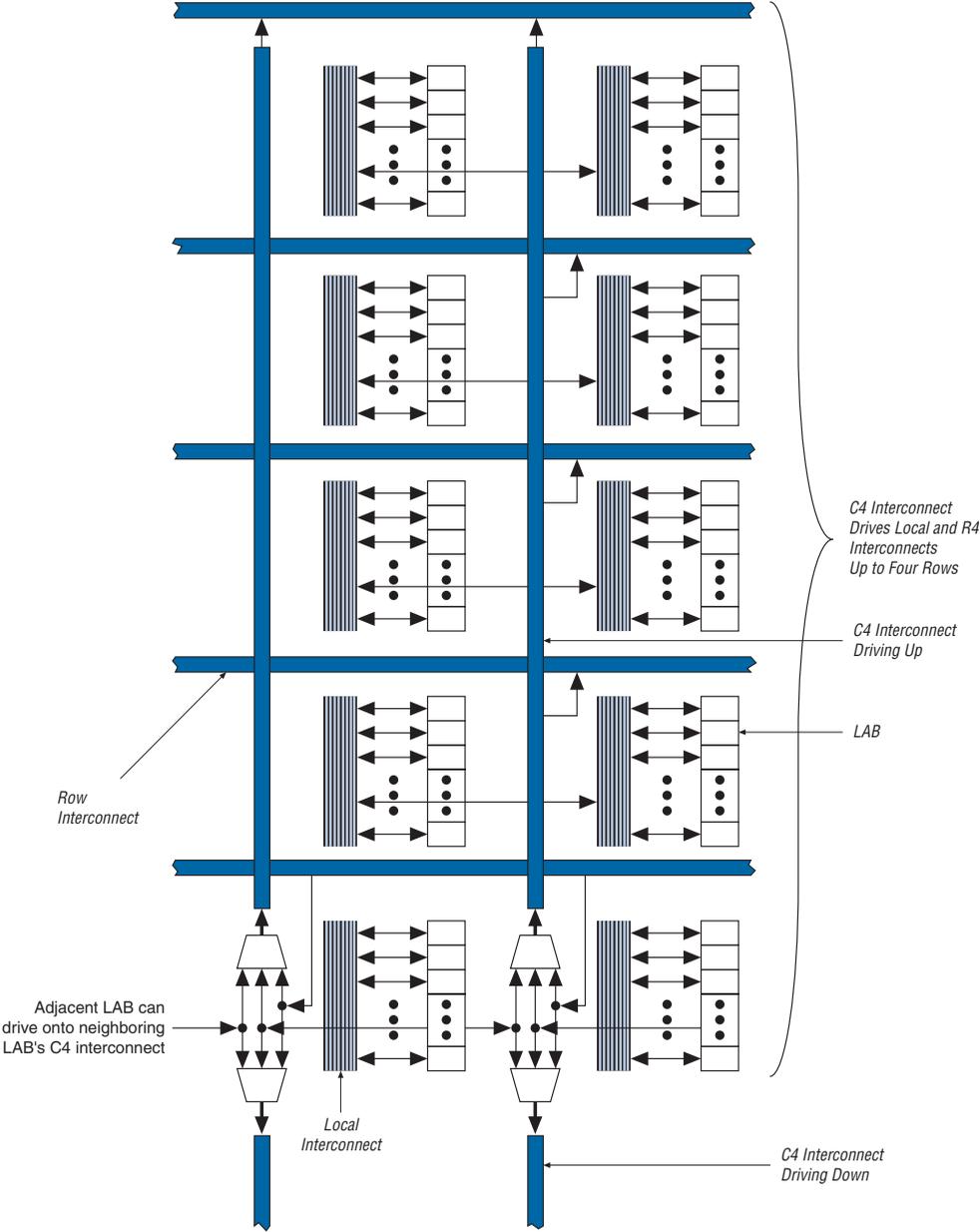
- LUT chain interconnects within an LAB
- Register chain interconnects within an LAB
- C4 interconnects traversing a distance of four LABs in an up and down direction

MAX II devices include an enhanced interconnect structure within LABs for routing LE output to LE input connections faster using LUT chain connections and register chain connections. The LUT chain connection allows the combinational output of an LE to directly drive the fast input of the LE right below it, bypassing the local interconnect. These resources can be used as a high-speed connection for wide fan-in functions from LE 1 to LE 10 in the same LAB. The register chain connection allows the register output of one LE to connect directly to the register input of the next LE in the LAB for fast shift registers. The Quartus II Compiler automatically takes advantage of these resources to improve utilization and performance. [Figure 2-11](#) shows the LUT chain and register chain interconnects.

**Figure 2–11. LUT Chain & Register Chain Interconnects**

The C4 interconnects span four LABs up or down from a source LAB. Every LAB has its own set of C4 interconnects to drive either up or down. [Figure 2–12](#) shows the C4 interconnect connections from an LAB in a column. The C4 interconnects can drive and be driven by column and row IOEs. For LAB interconnection, a primary LAB or its vertical LAB neighbor can drive a given C4 interconnect. C4 interconnects can drive each other to extend their range as well as drive row interconnects for column-to-column connections.

Figure 2-12. C4 Interconnect Connections Note (1)



Note to Figure 2-12:  
(1) Each C4 interconnect can drive either up or down four rows.

The UFM block communicates with the logic array similar to LAB-to-LAB interfaces. The UFM block connects to row and column interconnects and has local interconnect regions driven by row and column interconnects. This block also has DirectLink interconnects for fast connections to and from a neighboring LAB. For more information on the UFM interface to the logic array, see “User Flash Memory Block” on page 2–23.

Table 2–2 shows the MAX II device's routing scheme.

Source	Destination										
	LUT Chain	Register Chain	Local (1)	DirectLink (1)	R4 (1)	C4 (1)	LE	UFM Block	Column IOE	Row IOE	Fast I/O (1)
LUT Chain							✓				
Register Chain							✓				
Local Interconnect							✓	✓	✓	✓	
DirectLink Interconnect			✓								
R4 Interconnect			✓		✓	✓					
C4 Interconnect			✓		✓	✓					
LE	✓	✓	✓	✓	✓	✓			✓	✓	✓
UFM Block			✓	✓	✓	✓					
Column IOE						✓					
Row IOE				✓	✓	✓					

**Note to Table 2–2:**

(1) These categories are interconnects.

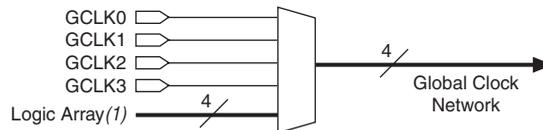
## Global Signals

Each MAX II device has four dual-purpose dedicated clock pins (GCLK[3..0]), two pins on the left side and two pins on the right side) that drive the global clock network for clocking, as shown in Figure 2–13. These four pins can also be used as general-purpose I/O if they are not used to drive the global clock network.

The four global clock lines in the global clock network drive throughout the entire device. The global clock network can provide clocks for all resources within the device including LEs, LAB local interconnect, IOEs, and the UFM block. The global clock lines can also be used for global

control signals, such as clock enables, synchronous or asynchronous clears, presets, output enables, or protocol control signals such as TRDY and IRDY for PCI. Internal logic can drive the global clock network for internally-generated global clocks and control signals. Figure 2-13 shows the various sources that drive the global clock network.

**Figure 2-13. Global Clock Generation**

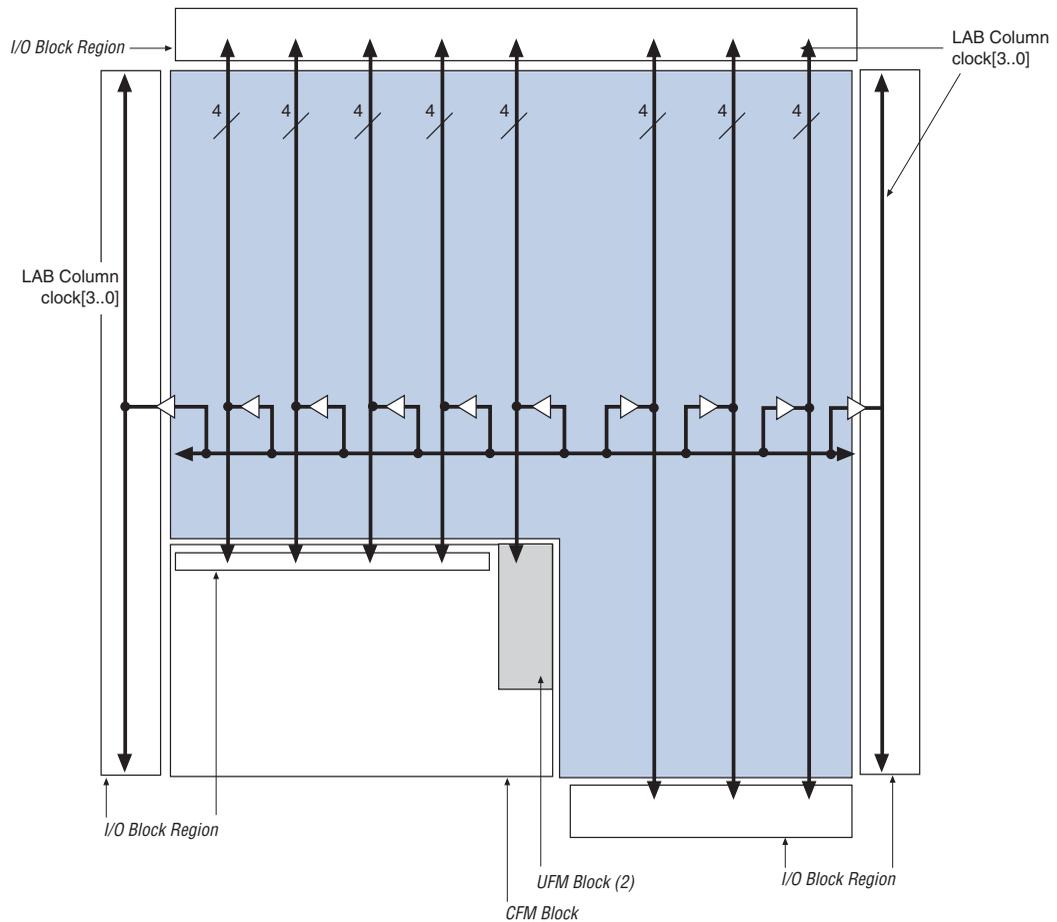


**Note to Figure 2-13:**

- (1) Any I/O pin can use a MultiTrack interconnect to route as a logic array-generated global clock signal.

The global clock network drives to individual LAB column signals, LAB column clocks [3..0], that span an entire LAB column from the top to bottom of the device. Unused global clocks or control signals in a LAB column are turned off at the LAB column clock buffers shown in Figure 2-14. The LAB column clocks [3..0] are multiplexed down to two LAB clock signals and one LAB clear signal. Other control signal types route from the global clock network into the LAB local interconnect. See “LAB Control Signals” on page 2-6 for more information.

**Figure 2-14. Global Clock Network** *Note (1)*



**Notes to Figure 2-14:**

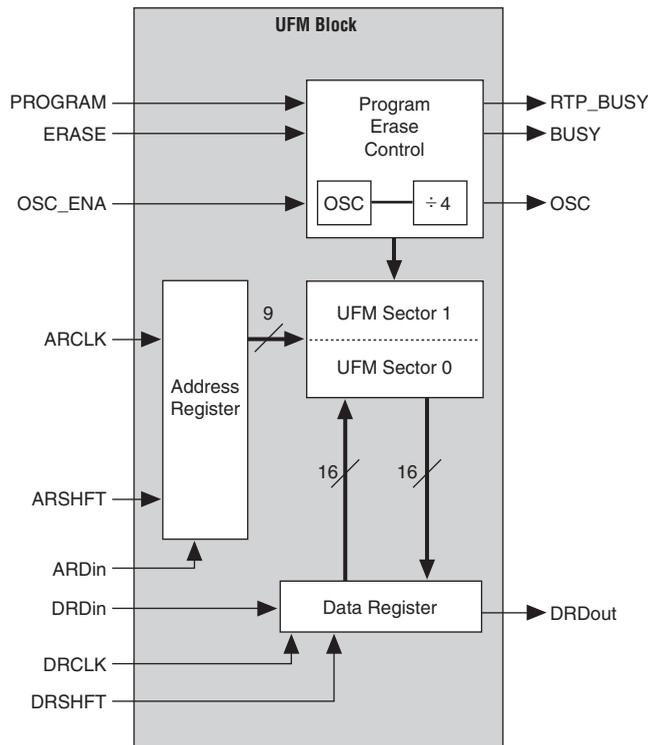
- (1) LAB column clocks in I/O block regions provide high fan-out output enable signals.
- (2) LAB column clocks drive to the UFM block.

## User Flash Memory Block

MAX II devices feature a single UFM block, which can be used like a serial EEPROM for storing non-volatile information up to 8,192 bits. The UFM block connects to the logic array through the MultiTrack interconnect, allowing any LE to interface to the UFM block. Figure 2–15 shows the UFM block and interface signals. The logic array is used to create customer interface or protocol logic to interface the UFM block data outside of the device. The UFM block offers the following features:

- Non-volatile storage up to 16-bit wide and 8,192 total bits
- Two sectors for partitioned sector erase
- Built-in internal oscillator that optionally drives logic array
- Program, erase, and busy signals
- Auto-increment addressing
- Serial interface to logic array with programmable interface

**Figure 2–15. UFM Block & Interface Signals**



## UFM Storage

Each device stores up to 8,192 bits of data in the UFM block. [Table 2–3](#) shows the data size, sector, and address sizes for the UFM block.

Device	Total Bits	Sectors	Address Bits	Data Width
EPM240 EPM570 EPM1270 EPM2210	8,192	2 (4,096 bits/sector)	9	16

There are 512 locations with 9-bit addressing ranging from 000h to 1FFh. Sector 0 address space is 000h to 0FFh and Sector 1 address space is from 100h to 1FFh. The data width is up to 16 bits of data. The Quartus II software automatically creates logic to accommodate smaller read or program data widths. Erasure of the UFM involves individual sector erasing (i.e., one erase of sector 0 and one erase of sector 1 is required to erase the entire UFM block). Since sector erase is required before a program or write, having two sectors enables a sector size of data to be left untouched while the other sector is erased and programmed with new data.

## Internal Oscillator

As shown in [Figure 2–15](#), the dedicated circuitry within the UFM block contains an oscillator. The dedicated circuitry uses this internally for its read and program operations. This oscillator's divide by 4 output can be driven out of the UFM block to the logic array for interface logic clock source or for general-purpose logic clocking. The OSC output signal frequency ranges from 4.6 to 7.4 MHz (preliminary), and its exact frequency is not programmable.

## Program, Erase & Busy Signals

The UFM block's dedicated circuitry automatically generates the necessary internal program and erase algorithm once the PROGRAM or ERASE input signals have been asserted. The PROGRAM or ERASE signal must be asserted until the busy signal deasserts, indicating the UFM internal program or erase operation has completed. The UFM block also supports JTAG as the interface for programming and/or reading.



For more information on programming and erasing the UFM block, see the chapter on *Using User Flash Memory in MAX II Devices*.

## Auto-Increment Addressing

The UFM block supports standard read or stream read operations. The stream read is supported with a auto-increment address feature. De-asserting the `ARSHIFT` signal while clocking the `ARCLK` signal increments the address register value to read consecutive locations from the UFM array.

## Serial Interface

The UFM block supports a serial interface with serial address and data signals. The internal shift registers within the UFM block for address and data are 9 bits and 16 bits wide, respectively. The Quartus II software automatically generates interface logic in LEs for a parallel address and data interface to the UFM block. Other standard protocol interfaces such as SPI are also automatically generated in LE logic by the Quartus II software.

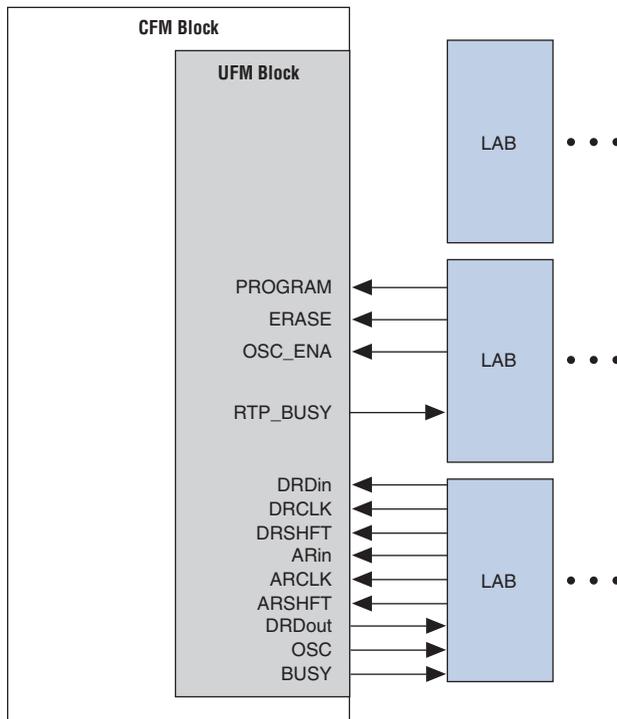


For more information on the UFM interface signals and the Quartus II LE-based alternate interfaces, see *Using User Flash Memory in MAX II Devices*.

## UFM Block to Logic Array Interface

The UFM block is a small partition of the flash memory which contains the CFM block as shown in [Figures 2-1](#) and [2-2](#). The UFM block for the EPM240 device is located on the left side of the device adjacent to the left most LAB column. The UFM block for the EPM570, EPM1270, and EPM2210 devices is located on the bottom left portion of the device. The UFM input and output signals interface to all types of interconnects (R4 interconnect, C4 interconnect, and DirectLink interconnect to/from adjacent LAB rows). The UFM signals can also be driven from global clocks, `GCLK[3..0]`. The interface region for the EPM240 device is shown in [Figure 2-16](#). The interface regions for EPM570, EPM1270, and EPM2210 devices are shown in [Figure 2-17](#).

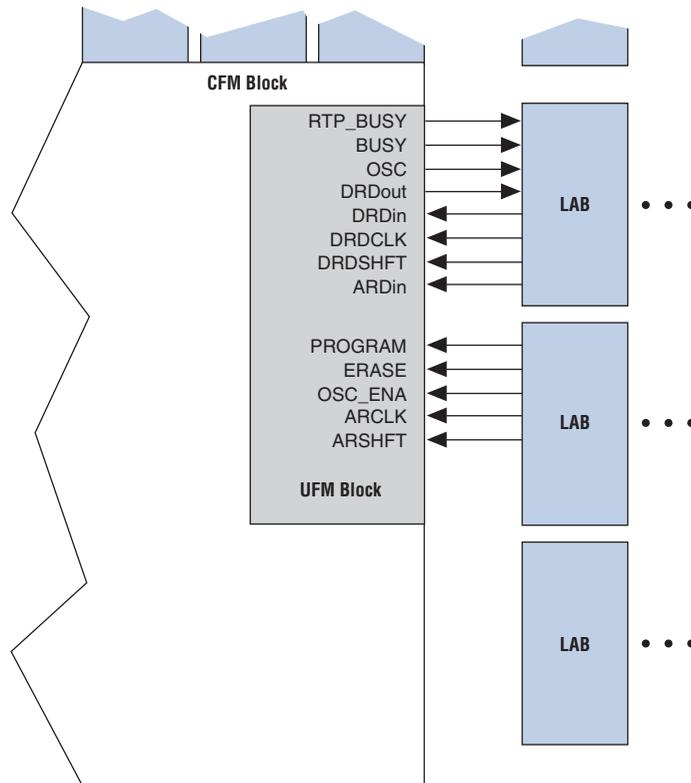
Figure 2-16. EPM240 UFM Block LAB Row Interface *Note (1)*



*Note to Figure 2-16:*

- (1) The UFM block inputs and outputs can drive to/from all types of interconnects, not only DirectLink interconnects from adjacent row LABs.

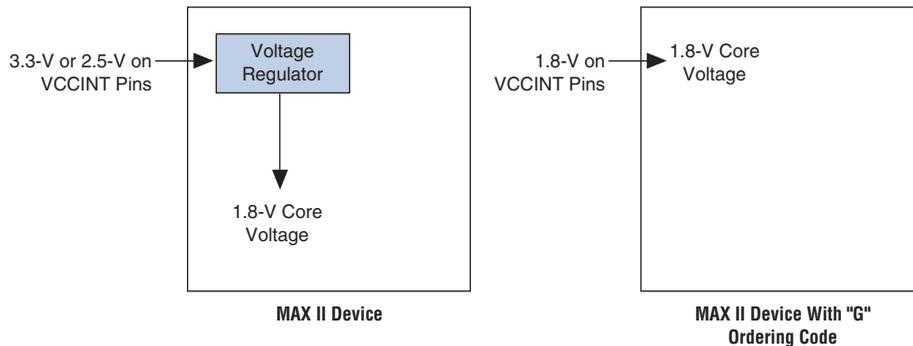
**Figure 2-17. EPM570, EPM1270 & EPM2210 UFM Block LAB Row Interface**



## MultiVolt Core

The MAX II architecture supports the MultiVolt™ core feature, which allows MAX II devices to support multiple  $V_{CC}$  levels on the  $V_{CCINT}$  supply. An internal linear voltage regulator provides the necessary 1.8-V internal voltage supply to the device. The voltage regulator supports 3.3-V or 2.5-V supplies on its inputs to supply the 1.8-V internal voltage to the device, as shown in Figure 2-18. The voltage regulator is not guaranteed for voltages that are between the maximum recommended 2.5-V operating voltage and the minimum recommended 3.3-V operating voltage.

For external 1.8-V supplies, MAX II devices with a “G” designation in the ordering code are required. The voltage regulator on these devices is bypassed to support the 1.8-V  $V_{CC}$  external supply path to the 1.8-V internal supply. Contact Altera for latest information regarding MAX II “G” devices.

**Figure 2–18. MultiVolt Core Feature in MAX II Devices**

## I/O Structure

IOEs support many features, including:

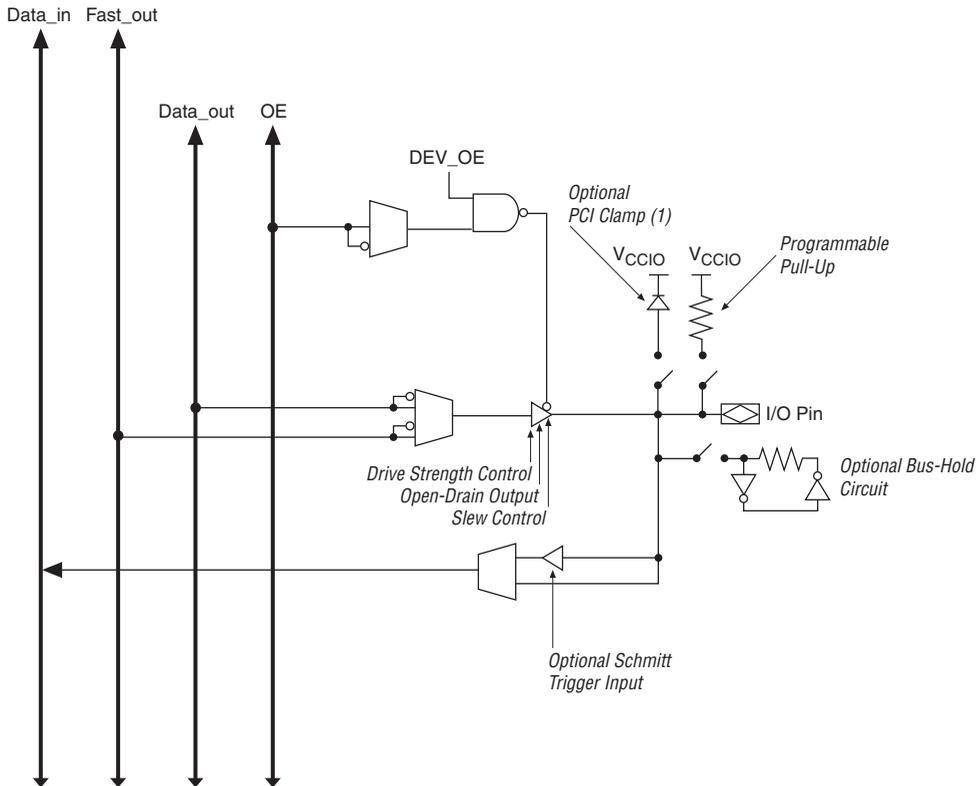
- LVTTTL and LVC MOS I/O standards
- 3.3-V, 32-bit, 33-MHz PCI compliance
- Joint Test Action Group (JTAG) boundary-scan test (BST) support
- Programmable drive strength control
- Weak pull-up resistors during power-up and in system programming
- Slew-rate control
- Tri-state buffers with individual output enable control
- Bus-hold circuitry
- Programmable pull-up resistors in user mode
- Unique output enable per pin
- Open-drain outputs
- Schmitt trigger inputs
- Fast I/O connection

MAX II device IOEs contain a bidirectional I/O buffer. [Figure 2–19](#) shows the MAX II IOE structure. Registers from adjacent LABs can drive to or be driven from the IOE's bidirectional I/O buffers. The Quartus II software automatically attempts to place registers in the adjacent LAB with fast I/O connection to achieve the fastest possible clock-to-output and registered output enable timing. For input registers, the Quartus II software automatically routes the register to guarantee zero hold time. You can set timing assignments in the Quartus II software to achieve desired I/O timing.

## Fast I/O Connection

A dedicated fast I/O connection from the adjacent LAB to the IOEs within an I/O block provides faster output delays for clock-to-output and  $t_{PD}$  propagation delays. This connection exists for data output signals, not output enable signals or input signals. Figures 2–20, 2–21, and 2–22 illustrate the fast I/O connection.

**Figure 2–19. MAX II IOE Structure**



**Note to Figure 2–19:**

(1) Available in EPM1270 and EPM2210 devices only.

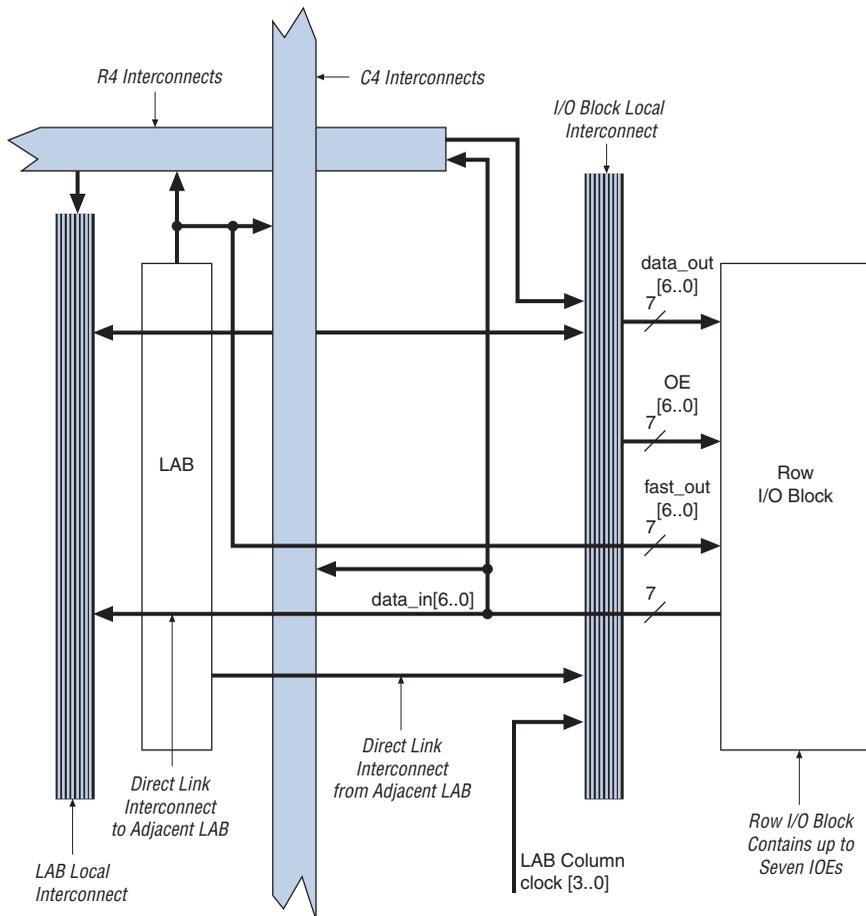
## I/O Blocks

The IOEs are located in I/O blocks around the periphery of the MAX II device. There are up to seven IOEs per row I/O block (5 maximum in the EPM240 device) and up to four IOEs per column I/O block. Each column or row I/O block interfaces with its adjacent LAB and MultiTrack

interconnect to distribute signals throughout the device. The row I/O blocks drive row, column, or DirectLink interconnects. The column I/O blocks drive column interconnects.

Figure 2–20 shows how a row I/O block connects to the logic array.

**Figure 2–20. Row I/O Block Connection to the Interconnect** Note (1)

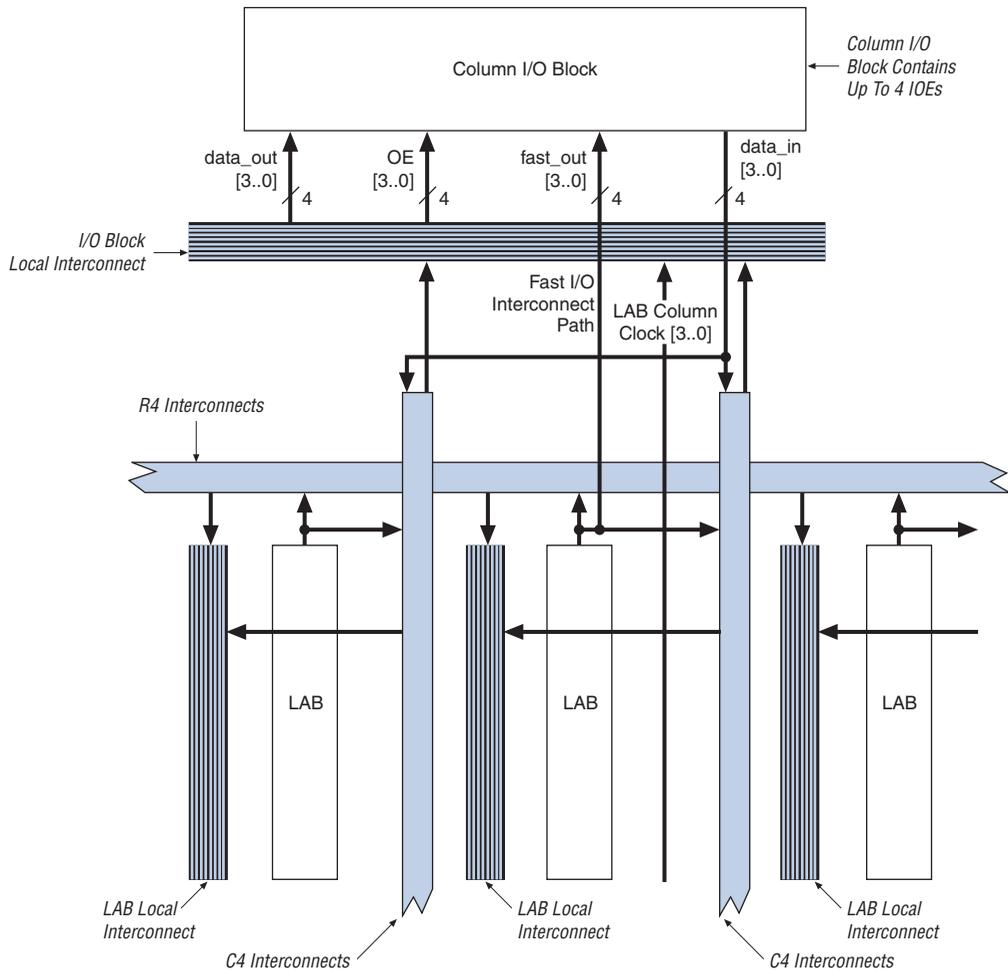


**Note to Figure 2–20:**

- (1) Each of the seven IOEs in the row I/O block can have one `data_out` or `fast_out` output, one `OE` output, and one `data_in` input.

Figure 2–21 shows how a column I/O block connects to the logic array.

**Figure 2–21. Column I/O Block Connection to the Interconnect** *Note (1)*



**Note to Figure 2–21:**

- (1) Each of the four IOEs in the column I/O block can have one data\_out or fast\_out output, one OE output, and one data\_in input.

## I/O Standards & Banks

MAX II device IOEs support the following I/O standards:

- 3.3-V LVTTTL/LVCMOS
- 2.5-V LVTTTL/LVCMOS
- 1.8-V LVTTTL/LVCMOS

- 1.5-V LVCMOS
- 3.3-V PCI

Table 2-4 describes the I/O standards supported by MAX II devices.

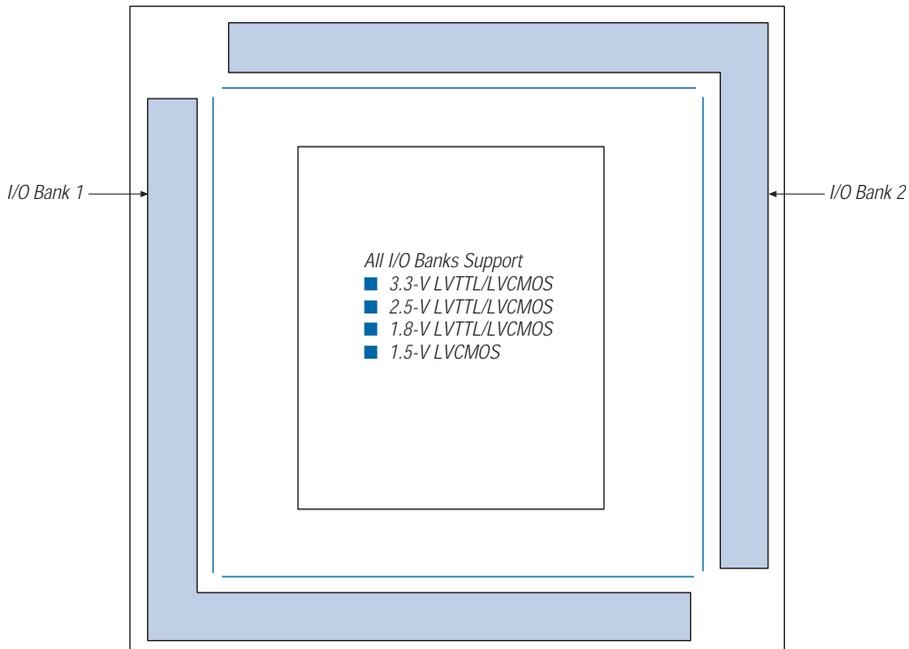
<b>I/O Standard</b>	<b>Type</b>	<b>Output Supply Voltage (<math>V_{CCIO}</math>) (V)</b>
3.3-V LVTTTL/LVCMOS	Single-ended	3.3
2.5-V LVTTTL/LVCMOS	Single-ended	2.5
1.8-V LVTTTL/LVCMOS	Single-ended	1.8
1.5-V LVCMOS	Single-ended	1.5
3.3-V PCI (1)	Single-ended	3.3

*Note to Table 2-4:*

- (1) 3.3-V PCI is supported in Bank 3 of the EPM1270 and EPM2210 devices.

The EPM240 and EPM570 devices support two I/O banks, as shown in Figure 2-22. Each of these banks support all the LVTTTL and LVCMOS standards shown in Table 2-4. PCI I/O is not supported in these devices and banks.

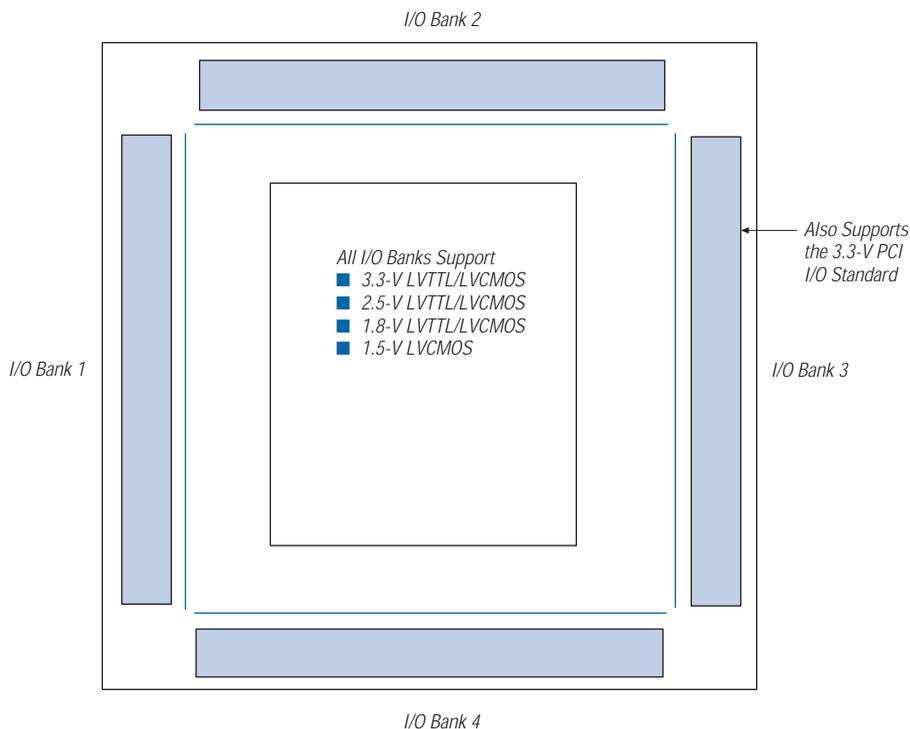
**Figure 2–22. MAX II I/O Banks for EPM240 & EPM570** Notes (1), (2)



**Notes to Figure 2–22:**

- (1) Figure 2–22 is a top view of the silicon die.
- (2) Figure 2–22 is a graphic representation only. Refer to the pin list and the Quartus II software for exact pin locations.

The EPM1270 and EPM2210 devices support four I/O banks, as shown in Figure 2–23. Each of these banks support all of the LVTTTL and LVCMOS standards shown in Table 2–4. PCI I/O is supported in Bank 3. Bank 3 supports the PCI clamping diode on inputs and PCI drive compliance on outputs. You must use Bank 3 for designs requiring PCI compliant I/O pins. The Quartus II software automatically places I/O pins in this bank if assigned with the PCI I/O standard.

**Figure 2–23. MAX II I/O Banks for EPM1270 & EPM2210** Notes (1), (2)**Notes to Figure 2–23:**

- (1) Figure 2–23 is a top view of the silicon die.
- (2) Figure 2–23 is a graphic representation only. Refer to the pin list and the Quartus II software for exact pin locations.

Each I/O bank has dedicated  $V_{CCIO}$  pins which determine the voltage standard support in that bank. A single device can support 1.5-V, 1.8-V, 2.5-V, and 3.3-V interfaces; each individual bank can support a different standard. Each I/O bank can support multiple standards with the same  $V_{CCIO}$  for input and output pins. For example, when  $V_{CCIO}$  is 3.3 V, Bank 3 can support LVTTTL, LVCMOS, and 3.3-V PCI.  $V_{CCIO}$  powers both the input and output buffers in MAX II devices.

The JTAG pins for MAX II devices are dedicated pins that cannot be used as regular I/O pins. The pins TMS, TDI, TDO, and TCK support all the I/O standards shown in Table 2–4 on page 2–32 except for PCI. These pins reside in Bank 1 for all MAX II devices and their I/O standard support is controlled by the  $V_{CCIO}$  setting for Bank 1.

### PCI Compliance

MAX II EPM1270 and EPM2210 devices are compliant with PCI applications as well as all 3.3-V electrical specifications in the *PCI Local Bus Specification Revision 2.2*. These devices are also large enough to support PCI intellectual property (IP) cores. [Table 2-5](#) shows the MAX II device speed grades that meet the PCI timing specifications.

**Table 2-5. MAX II Devices & Speed Grades that Support 3.3-V PCI Electrical Specifications & Meet PCI Timing** *Note (1)*

Device	33-MHz PCI
EPM1270	All Speed Grades
EPM2210	All Speed Grades

*Note to Table 2-5:*

(1) This table contains preliminary information.

### Schmitt Trigger

The input buffer for each MAX II device I/O pin has an optional Schmitt trigger setting for the 3.3-V and 2.5-V standards. The Schmitt trigger allows input buffers to respond to slow input edge rates with a fast output edge rate. Most importantly, Schmitt triggers provide hysteresis on the input buffer, preventing slow rising noisy input signals from ringing or oscillating on the input signal driven into the logic array. This provides system noise tolerance on MAX II inputs, but adds a small, nominal input delay.

The JTAG input pins (TMS, TCK, and TDI) have Schmitt trigger buffers which are always enabled.

### Output Enable Signals

Each MAX II IOE output buffer supports output enable signals for tri-state control. The output enable signal can originate from the GCLK[3..0] global signals or from the MultiTrack interconnect. The MultiTrack interconnect routes output enable signals and allows for a unique output enable for each output or bidirectional pin.

MAX II devices also provide a chip-wide output enable pin (DEV\_OE) to control the output enable for every output pin in the design. An option set before compilation in the Quartus II software controls this pin. If this option is turned on, all outputs on the chip operate normally when DEV\_OE is asserted. When the pin is de-asserted, all outputs are tri-stated. If this option is turned off, the DEV\_OE pin is disabled when the device operates in user mode and is available as a user I/O pin.

## Programmable Drive Strength

The output buffer for each MAX II device I/O pin has two levels of programmable drive strength control for each of the LVTTTL and LVCMOS I/O standards. Programmable drive strength provides system noise reduction control for high performance I/O designs. Although a separate slew-rate control feature exists, using the lower drive strength setting provides signal slew rate control to reduce system noise and signal overshoot without the large delay adder associated with the slew-rate control feature. Table 2-6 shows the possible settings for the I/O standards with drive strength control. The PCI I/O standard is always set at 20 mA with no alternate setting.

I/O Standard	I <sub>OH</sub> /I <sub>OL</sub> Current Strength Setting (mA)
3.3-V LVTTTL	16
	8
3.3-V LVCMOS	8
	4
2.5-V LVTTTL/LVCMOS	14
	7
1.8-V LVTTTL/LVCMOS	6
	3
1.5-V LVCMOS	4
	2

**Note to Table 2-6:**

- (1) The I<sub>OH</sub> current strength numbers shown are for a condition of a V<sub>OUT</sub> = V<sub>OH</sub> minimum, where the V<sub>OH</sub> minimum is specified by the I/O standard. The I<sub>OL</sub> current strength numbers shown are for a condition of a V<sub>OUT</sub> = V<sub>OL</sub> maximum, where the V<sub>OL</sub> maximum is specified by the I/O standard. For 2.5-V LVTTTL/LVCMOS, the I<sub>OH</sub> condition is V<sub>OUT</sub> = 1.7 V and the I<sub>OL</sub> condition is V<sub>OUT</sub> = 0.7 V.

## Slew-Rate Control

The output buffer for each MAX II device I/O pin has a programmable output slew-rate control that can be configured for low noise or high-speed performance. A faster slew rate provides high-speed transitions for high-performance systems. However, these fast transitions may introduce noise transients into the system. A slow slew rate reduces system noise, but adds a nominal output delay to rising and falling edges. The lower the voltage standard (e.g., 1.8-V LVTTTL) the larger the output delay when slow slew is enabled. Each I/O pin has an individual slew-rate control, allowing the designer to specify the slew rate on a pin-by-pin basis. The slew-rate control affects both the rising and falling edges.

## Open-Drain Output

MAX II devices provide an optional open-drain (equivalent to open-collector) output for each I/O pin. This open-drain output enables the device to provide system-level control signals (e.g., interrupt and write enable signals) that can be asserted by any of several devices. This output can also provide an additional wired-OR plane.

## Programmable Ground Pins

Each unused I/O pin on MAX II devices can be used as an additional ground pin. This programmable ground feature does not require the use of the associated LEs in the device. In the Quartus II software, unused pins can be set as programmable GND on a global default basis or they can be individually assigned. Unused pins also have the option of being set as tri-stated input pins.

## Bus Hold

Each MAX II device I/O pin provides an optional bus-hold feature. The bus-hold circuitry can hold the signal on an I/O pin at its last-driven state. Since the bus-hold feature holds the last-driven state of the pin until the next input signal is present, an external pull-up or pull-down resistor is not necessary to hold a signal level when the bus is tri-stated.

The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. The designer can select this feature individually for each I/O pin. The bus-hold output will drive no higher than  $V_{CCIO}$  to prevent overdriving signals. If the bus-hold feature is enabled, the device cannot use the programmable pull-up option.

The bus-hold circuitry uses a resistor with a nominal resistance (RBH) of approximately 5 to 40 k $\Omega$  to pull the signal level to the last driven state. The chapter on *DC & Switching Characteristics* gives the specific sustaining current for each V<sub>CCIO</sub> voltage level driven through this resistor and overdrive current used to identify the next-driven input level.

The bus-hold circuitry is only active after the device has fully initialized. The bus-hold circuit captures the value on the pin present at the moment user mode is entered.

### Programmable Pull-Up Resistor

Each MAX II device I/O pin provides an optional programmable pull-up resistor during user mode. If the designer enables this feature for an I/O pin, the pull-up resistor (approximately 5 to 40 k $\Omega$ ) holds the output to the V<sub>CCIO</sub> level of the output pin's bank.



The programmable pull-up resistor feature should not be used at the same time as the bus-hold feature on a given I/O pin.

### MultiVolt I/O Interface

The MAX II architecture supports the MultiVolt I/O interface feature, which allows MAX II devices in all packages to interface with systems of different supply voltages. The devices have one set of VCC pins for internal operation (VCCINT), and four sets for input buffers and I/O output driver buffers (VCCIO).

Connect V<sub>CCIO</sub> pins to either a 1.5-V, 1.8 V, 2.5-V, or 3.3-V power supply, depending on the output requirements. The output levels are compatible with systems of the same voltage as the power supply (i.e., when V<sub>CCIO</sub> pins are connected to a 1.5-V power supply, the output levels are compatible with 1.5-V systems). When V<sub>CCIO</sub> pins are connected to a 3.3-V power supply, the output high is 3.3 V and is compatible with 3.3-V or 5.0-V systems. Table 2–7 summarizes MAX II MultiVolt I/O support.

V <sub>CCIO</sub> (V)	Input Signal					Output Signal				
	1.5 V	1.8 V	2.5 V	3.3 V	5.0 V	1.5 V	1.8 V	2.5 V	3.3 V	5.0 V
1.5	✓	✓	✓ (2)	✓ (2)		✓				
1.8		✓	✓	✓		✓ (3)	✓			
2.5			✓	✓		✓ (4)	✓ (4)	✓		
3.3			✓ (5)	✓	✓ (6)	✓ (7)	✓ (7)	✓ (7)	✓	✓ (8)

**Notes to Table 2–7:**

- (1) To drive inputs higher than V<sub>CCIO</sub> but less than 4.0 V including the overshoot, disable the PCI clamping diode. However, to drive 5.0-V inputs to the device, enable the PCI clamping diode to prevent V<sub>I</sub> from rising above 4.0 V.
- (2) When V<sub>CCIO</sub> = 1.5-V and a 2.5-V or 3.3-V input signal feeds an input pin, higher pin leakage current is expected.
- (3) When V<sub>CCIO</sub> = 1.8-V, a MAX II device can drive a 1.5-V device with 1.8-V tolerant inputs.
- (4) When V<sub>CCIO</sub> = 2.5-V, a MAX II device can drive a 1.5-V or 1.8-V device with 2.5-V tolerant inputs.
- (5) When V<sub>CCIO</sub> = 3.3-V and a 2.5-V input signal feeds an input pin, the V<sub>CCIO</sub> supply current will be slightly larger than expected.
- (6) MAX II devices can be 5.0-V tolerant with the use of an external resistor and the internal PCI clamp diode on the EPM1270 and EPM2210 devices.
- (7) When V<sub>CCIO</sub> = 3.3-V, a MAX II device can drive a 1.5-V, 1.8-V, or 2.5-V device with 3.3-V tolerant inputs.
- (8) When V<sub>CCIO</sub> = 3.3-V, a MAX II device can drive a device with 5.0-V TTL inputs but not 5.0-V CMOS inputs. In the case of 5.0-V CMOS, open-drain setting with internal PCI clamp diode (available only on EPM1270 and EPM2210 devices) and external resistor is required.



## IEEE Std. 1149.1 (JTAG) Boundary Scan Support

All MAX<sup>®</sup> II devices provide Joint Test Action Group (JTAG) boundary-scan test (BST) circuitry that complies with the IEEE Std. 1149.1-2001 specification. JTAG boundary-scan testing can only be performed at any time after  $V_{CCINT}$  and all  $V_{CCIO}$  banks have been fully powered and a  $t_{CONFIG}$  amount of time has passed. MAX II devices can also use the JTAG port for in-system programming together with either the Quartus<sup>®</sup> II software or hardware using Programming Object Files (.pof), Jam<sup>™</sup> Standard Test and Programming Language (STAPL) Files (.jam) or Jam Byte-Code Files (.jbc).

The JTAG pins support 1.5-V, 1.8-V, 2.5-V, or 3.3-V I/O standards. The supported voltage level and standard is determined by the  $V_{CCIO}$  of the bank where it resides. The dedicated JTAG pins reside in Bank 1 of all MAX II devices.

MAX II devices support the JTAG instructions shown in [Table 3-1](#).

**Table 3-1. MAX II JTAG Instructions (Part 1 of 2)**

JTAG Instruction	Instruction Code	Description
SAMPLE/PRELOAD	00 0000 0101	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.
EXTEST (1)	00 0000 1111	Allows the external circuitry and board-level interconnects to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	11 1111 1111	Places the 1-bit bypass register between the TDI and TDO pins, which allows the boundary scan test data to pass synchronously through selected devices to adjacent devices during normal device operation.
USERCODE	00 0000 0111	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO. This register defaults to all 1's if not specified in the Quartus II software.
IDCODE	00 0000 0110	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.

**Table 3–1. MAX II JTAG Instructions (Part 2 of 2)**

JTAG Instruction	Instruction Code	Description
HIGHZ (1)	00 0000 1011	Places the 1-bit bypass register between the TDI and TDO pins, which allows the boundary scan test data to pass synchronously through selected devices to adjacent devices during normal device operation, while tri-stating all of the I/O pins.
CLAMP (1)	00 0000 1010	Places the 1-bit bypass register between the TDI and TDO pins, which allows the boundary scan test data to pass synchronously through selected devices to adjacent devices during normal device operation, while holding I/O pins to a state defined by the data in the boundary-scan register.
USER0	00 0000 1100	This instruction allows the user to define their own scan chain between TDI and TDO in the MAX II logic array. This instruction is also used for custom logic and JTAG interfaces.
USER1	00 0000 1110	This instruction allows the user to define their own scan chain between TDI and TDO in the MAX II logic array. This instruction is also used for custom logic and JTAG interfaces.
IEEE 1532 instructions	(2)	IEEE 1532 ISC instructions used when programming a MAX II device via the JTAG port.

**Notes to Table 3–1:**

- (1) HIGHZ, CLAMP, and EXTEST instructions do not disable weak pull-up resistors or bus hold features.
- (2) These instructions are shown in the 1532 BSDL files, which will be posted on the Altera® web site at [www.altera.com](http://www.altera.com) when they are available.

The MAX II device instruction register length is 10 bits and the USERCODE register length is 32 bits. Tables 3-2 and 3-3 show the boundary-scan register length and device IDCODE information for MAX II devices.

**Table 3-2. MAX II Boundary-Scan Register Length**

Device	Boundary-Scan Register Length
EPM240	240
EPM570	480
EPM1270	636
EPM2210	816

**Table 3-3. 32-Bit MAX II Device IDCODE**

Device	Binary IDCODE (32 Bits) (1)				HEX IDCODE
	Version (4 Bits)	Part Number	Manufacturer Identity (11 Bits)	LSB (1 Bit) (2)	
EPM240	0000	0010 0000 1010 0001	000 0110 1110	1	0x020A10DD
EPM570	0000	0010 0000 1010 0010	000 0110 1110	1	0x020A20DD
EPM1270	0000	0010 0000 1010 0011	000 0110 1110	1	0x020A30DD
EPM2210	0000	0010 0000 1010 0100	000 0110 1110	1	0x020A40DD

**Notes to Table 3-2:**

- (1) The most significant bit (MSB) is on the left.
- (2) The IDCODE's least significant bit (LSB) is always 1.



For JTAG AC characteristics, refer to the chapter on *DC & Switching Characteristics*. For more information on JTAG BST, see the chapter on *IEEE 1149.1 (JTAG) Boundary-Scan Testing for MAX II Devices*.

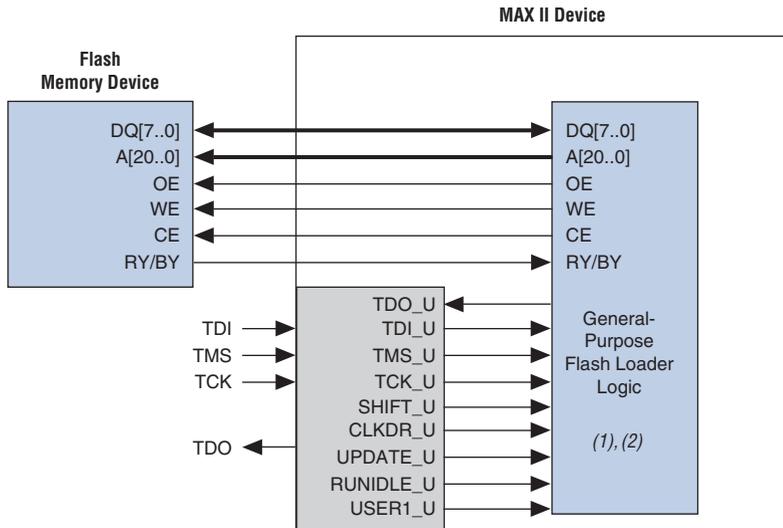
### JTAG Translator

The JTAG translator feature allows you to access the JTAG TAP and state signals when either the USER0 or USER1 instruction is issued to the JTAG TAP. The USER0 and USER1 instructions bring the JTAG boundary scan chain (TDI) through the user logic instead of the MAX II device's boundary scan cells. Each USER instruction allows for one unique user-defined JTAG chain into the logic array.

### General-Purpose Flash Loader

The JTAG translator ability to interface JTAG to non-JTAG devices is ideal for general-purpose flash memory devices (such as Intel or Fujitsu based devices) that require programming during in-circuit test. The flash memory devices can be used for FPGA configuration or be part of system memory. In many cases, the MAX II device is already connected to these devices as the configuration control logic between the FPGA and the flash device. Unlike ISP-capable CPLD devices, bulk flash devices do not have JTAG TAP pins or connections. For small flash devices, it is common to use the serial JTAG scan chain of a connected device to program the non-JTAG flash device. This is slow and inefficient in most cases and impractical for large parallel flash devices. Using the MAX II device's JTAG translator as a general-purpose flash loader to program and verify flash contents provides a fast and cost-effective means of in-circuit programming during test. Figure 3-1 shows MAX II being used as a general-purpose flash loader.

Figure 3-1. MAX II JTAG Translator as General-Purpose Flash Loader



Notes to Figure 3-1:

- (1) This block is implemented in LEs.
- (2) This function will be supported in a future version of the Quartus II software.

## In System Programmability

MAX II devices can be programmed in-system via the industry standard 4-pin IEEE Std. 1149.1 (JTAG) interface. In system programmability (ISP) offers quick, efficient iterations during design development and

debugging cycles. The logic, circuitry, and interconnects in the MAX II architecture are configured with flash-based SRAM configuration elements. These SRAM elements require configuration data to be loaded each time the device is powered. The process of loading the SRAM data is called configuration. The on-chip configuration flash memory (CFM) block stores the SRAM element's configuration data. The CFM block stores the design's configuration pattern in a reprogrammable flash array. During ISP, the MAX II JTAG and ISP circuitry programs the design pattern into the CFM block's non-volatile flash array.

The MAX II JTAG and ISP controller internally generate the high programming voltages required to program the CFM cells, allowing in-system programming with any of the recommended operating external voltage supplies (i.e., 3.3 V/2.5 V or 1.8 V for the MAX II devices with a "G" ordering code). ISP can be performed anytime after  $V_{CCINT}$  and all  $V_{CCIO}$  banks have been fully powered and the device has completed the configuration power-up time. By default, during in-system programming, the I/O pins are tri-stated and weakly pulled-up to  $V_{CCIO}$  to eliminate board conflicts. The pull-up value ranges from 5 to 40 k $\Omega$ . There are two other options in MAX II devices that allow user control of I/O state or behavior during ISP.



For more information, refer to ["In-System Programming Clamp"](#) on page 3–8 and ["Real-Time ISP"](#) on page 3–8.

These devices also offer an `ISP_DONE` bit that provides safe operation when in-system programming is interrupted. This `ISP_DONE` bit, which is the last bit programmed, prevents all I/O pins from driving until the bit is programmed.

## IEEE 1532 Support

The JTAG circuitry and ISP instruction set in MAX II devices is compliant to the IEEE 1532-2002 programming specification. This provides industry-standard hardware and software for in-system programming among multiple vendor programmable logic devices (PLDs) in a JTAG chain.

The MAX II 1532 BSDL files will be released on the Altera web site when available.

## Jam Standard Test & Programming Language (STAPL)

The Jam STAPL JEDEC standard, JESD71, can be used to program MAX II devices with in-circuit testers, PCs, or embedded processors. The Jam byte code is also supported for MAX II devices. These software programming protocols provide a compact embedded solution for programming MAX II devices.



For more information, see the chapter on *Using Jam STAPL for ISP via an Embedded Processor*.

## Programming Sequence

During in-system programming, 1532 instructions, addresses, and data are shifted into the MAX II device through the TDI input pin. Data is shifted out through the TDO output pin and compared against the expected data. Programming a pattern into the device requires the following six ISP steps. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6. These steps are automatically executed by third-party programmers, the Quartus® II software, or the Jam STAPL and Jam Byte-Code Players.

1. *Enter ISP* – The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode.
2. *Check ID* – Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
3. *Bulk Erase* – Erasing the device in-system involves shifting in the instruction to erase the device and applying an erase pulse(s). The erase pulse is automatically generated internally by waiting in the run/test/idle state for the specified erase pulse time of 350 ms.
4. *Program* – Programming the device in-system involves shifting in the address, data, and program instruction and generating the program pulse to program the flash cells. The program pulse is automatically generated internally by waiting in the run/test/idle state for the specified program pulse time of 75  $\mu$ s. This process is repeated for each address in the CFM block.
5. *Verify* – Verifying a MAX II device in-system involves shifting in addresses, applying the verify instruction to generate the read pulse, and shifting out the data for comparison. This process is repeated for each CFM address.
6. *Exit ISP* – An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode.

For TCK frequencies of 10 MHz, the erase and programming takes less than one second for EPM240 and EPM570 devices. Erase and programming times are less than two seconds for EPM1270 and less than three seconds for the EPM2210 devices. The TCK frequency can operate at up to 25 MHz in MAX II devices providing slight improvements in these ISP times.

### UFM Programming

The Quartus II software, with the use of POF, Jam, or JBC files, supports programming of each user flash memory (UFM) block sector independent from the logic array design pattern stored in the CFM block. This allows updating or reading UFM contents through ISP without altering the current logic array design, or vice versa. By default, these programming files and methods will program both the entire flash memory contents, which includes the CFM block and UFM contents. The stand-alone embedded Jam STAPL player and Jam Byte-Code Player provides action commands for programming or reading the entire flash memory (UFM and CFM together) or each independently.



For more information, see the chapter on *Using Jam STAPL for ISP via an Embedded Processor*.

### In-System Programming Clamp

By default, the IEEE 1532 instruction used for entering ISP automatically tri-states all I/O pins with weak pull-up resistors for the duration of the ISP sequence. However, some systems may require certain pins on MAX II devices to maintain a specific DC logic level during an in-field update. For these systems, an optional in-system programming clamp instruction exists in MAX II circuitry to control I/O behavior during the ISP sequence. The in-system programming clamp instruction enables the device to sample and sustain the value on an output pin (an input pin would remain tri-stated if sampled) or to explicitly set a logic high, logic low, or tri-state value on any pin. Setting these options is controlled on an individual pin basis using the Quartus II software.

### Real-Time ISP

For systems that require more than DC logic level control of I/O pins, the real-time ISP feature allows you to update the CFM block with a new design image while the entire current design continues to operate in the SRAM logic array and I/O pins. A new programming file is updated into the MAX II device without halting the original design's operation, saving down-time costs for remote or field upgrades. The updated CFM block configures the new design into the SRAM upon the next power cycle. It is also possible to execute an immediate configuration of the SRAM without a power cycle by using a specific sequence of ISP commands. The configuration of SRAM without a power cycle takes a specific amount of time ( $t_{\text{CONFIG}}$ ). During this time, the I/O pins are tri-stated and weakly pulled-up to  $V_{\text{CCIO}}$ .

The Quartus II software, Jam STAPL player, and Jam Byte-Code Player provides the designer with the option to control real-time ISP (if used) and allows the designer to control the timeline of new design configuration into the SRAM (immediately or upon next power cycle).

## **Design Security**

All MAX II devices contain a programmable security bit that controls access to the data programmed into the CFM block. When this bit is programmed, design programming information, stored in the CFM block, cannot be copied or retrieved. This feature provides a high level of design security because programmed data within flash memory cells is invisible. The security bit that controls this function, as well as all other programmed data, is reset only when the device is erased. The SRAM is also invisible and cannot be accessed regardless of the security bit setting. The UFM block data is not protected by the security bit and is accessible through JTAG or logic array connections.

## **Programming with External Hardware**

MAX II devices can be programmed by downloading the information via in-circuit testers, embedded processors, the Altera® ByteblasterMV™, MasterBlaster™, ByteBlaster™ II, and USB-Blaster cables, and through the universal serial bus (USB)-based Altera Programming Unit (APU) with the appropriate adapter.

BP Microsystems, System General, and other programming hardware manufacturers provide programming support for Altera devices. Check their web sites for device support information.



## Hot Socketing

MAX<sup>®</sup> II devices offer hot socketing, also known as hot plug-in or hot swap, and power sequencing support. Designers can insert or remove a MAX II board in a system during operation without undesirable effects to the system bus. The hot socketing feature removes some of the difficulty designers face when using components on printed circuit boards (PCBs) that contain a mixture of 3.3-, 2.5-, 1.8-, and 1.5-V devices.

The MAX II device hot socketing feature provides:

- Board or device insertion and removal
- Support for any power-up sequence
- Non-intrusive I/O buffers to system buses during hot insertion

### MAX II Hot-Socketing Specifications

MAX II devices offer all three of the features required for hot socketing capability listed above without any external components or special design requirements. The following are hot-socketing specifications:

- The device can be driven before and during power-up or power-down without any damage to the device itself.
- I/O pins remain tri-stated during power-up. The device does not drive out before or during power-up, thereby affecting other buses in operation.
- Signal pins do not drive the  $V_{CCIO}$  or  $V_{CCINT}$  power supplies. External input signals to I/O pins of the device do not internally power the  $V_{CCIO}$  or  $V_{CCINT}$  power supplies of the device via internal paths.

#### *Devices Can Be Driven before Power-Up*

Signals can be driven into the MAX II device I/O pins and  $GCLK[3..0]$  pins before or during power-up or power-down without damaging the device. MAX II devices support any power-up or power-down sequence ( $V_{CCIO1}$ ,  $V_{CCIO2}$ ,  $V_{CCIO3}$ ,  $V_{CCIO4}$ ,  $V_{CCINT}$ ), simplifying system-level design.

### *I/O Pins Remain Tri-States during Power-Up*

A device that does not support hot-socketing may interrupt system operation or cause contention by driving out before or during power-up. In a hot socketing situation, the MAX II device's output buffers are turned off during system power-up. MAX II devices do not drive out until the device attains proper operating conditions and is fully configured. See [“Power-On Reset Circuitry” on page 4–6](#) for information about turn-on voltages.

### *Signal Pins Do Not Drive the $V_{CCIO}$ or $V_{CCINT}$ Power Supplies*

MAX II devices do not have a current path from I/O pins or  $GCLK[3..0]$  pins to the  $V_{CCIO}$  or  $V_{CCINT}$  pins before or during power-up. A MAX II device may be inserted into (or removed from) a system board that was powered up without damaging or interfering with system-board operation. When hot socketing, MAX II devices may have a minimal effect on the signal integrity of the backplane.

### *AC & DC Specifications*

You can power up or power down the  $V_{CCIO}$  and  $V_{CCINT}$  pins in any sequence. There are no  $V_{CC}$  ramp rate requirements for MAX II devices. During hot socketing, the I/O pin capacitance is less than 8 pF. MAX II devices meet the following hot socketing specification:

- The hot socketing DC specification is  $|I_{IOPIN}| < 300 \mu\text{A}$ .
- The hot socketing AC specification is dependent on the signal voltages and board capacitance:  $|I_{IOPIN}| < (\Delta V / \Delta t) \times \text{capacitance}$  where capacitance is the sum of I/O pin, trace, and connector capacitance.



MAX II devices are immune to latch-up when hot socketing.

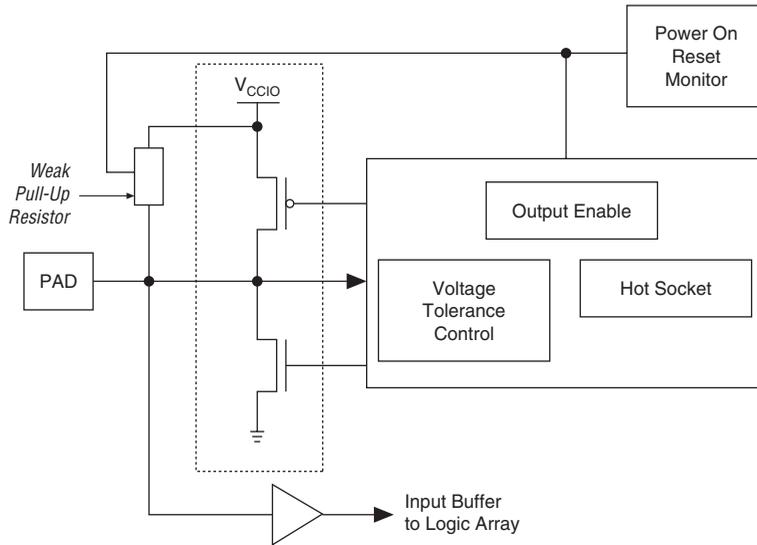
If the TCK JTAG input pin is driven high during hot-socketing, the current on that pin might exceed the specifications above.

### **Hot Socketing Feature Implementation in MAX II Devices**

The hot socketing feature turns off (tri-states) the output buffer during the power-up event (either  $V_{CCINT}$  or  $V_{CCIO}$  supplies) or power down. The hot-socket circuit generates an internal  $HOTSCKT$  signal when either  $V_{CCINT}$  or  $V_{CCIO}$  is below the threshold voltage. The  $HOTSCKT$  signal cuts off the output buffer to make sure that no DC current (except for weak pull-up leaking) leaks through the pin. When  $V_{CC}$  ramps up very slowly,  $V_{CC}$  may still be relatively low even after the power-on reset (POR) signal is released and device configuration is complete.

Each I/O and clock pin has the following circuitry, as shown in Figure 4-1.

**Figure 4-1. Hot Socketing Circuit Block Diagram for MAX II Devices**



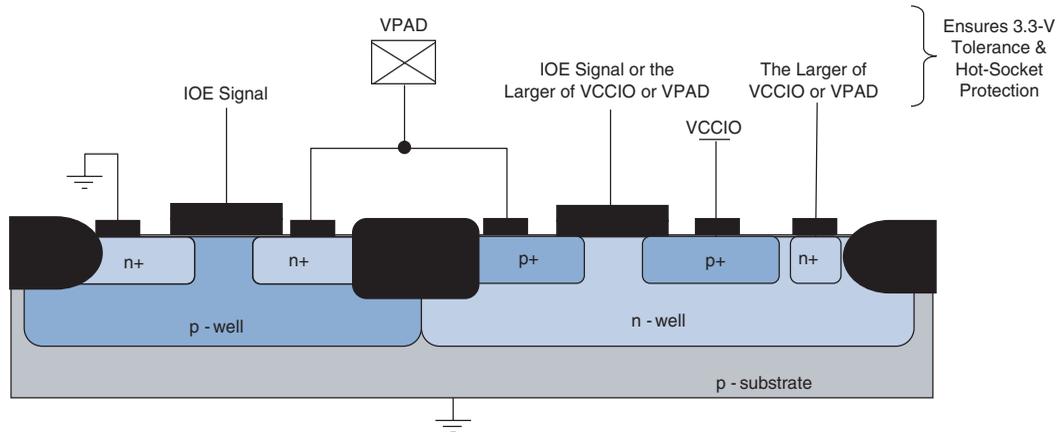
The POR circuit monitors  $V_{CCINT}$  and  $V_{CCIO}$  voltage levels and keeps I/O pins tri-stated until the device has completed its flash memory configuration of the SRAM logic. The weak pull-up resistor (R) from the I/O pin to  $V_{CCIO}$  is present to keep the I/O pins from floating. The 3.3-V tolerance control circuit permits the I/O pins to be driven by 3.3 V before  $V_{CCIO}$  and/or  $V_{CCINT}$  are powered, and it prevents the I/O pins from driving out when the device is not fully powered or operational. The hot-socket circuit prevents I/O pins from internally powering  $V_{CCIO}$  and  $V_{CCINT}$  when driven by external signals before the device is powered.



For information on 5.0-V tolerance, See the chapter on *Using MAX II Devices in Multi-Voltage Systems*.

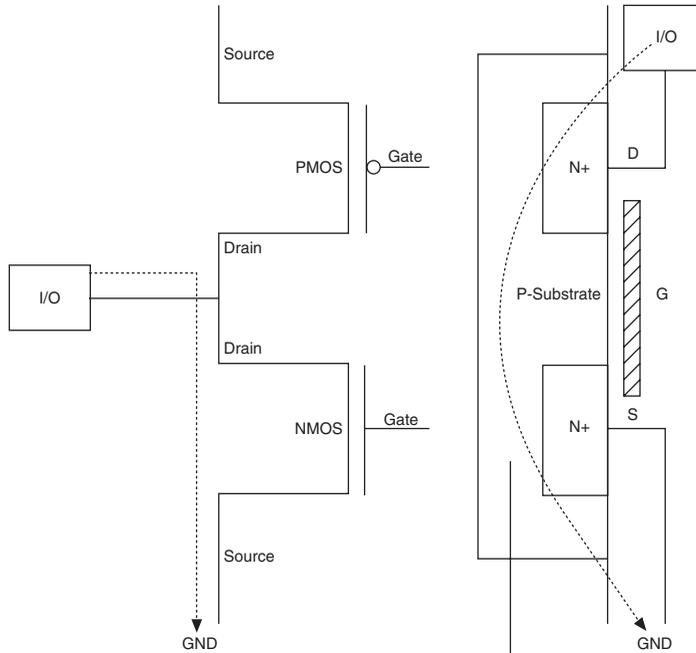
Figure 4-2 shows a transistor level cross section of the MAX II device I/O buffers. This design ensures that the output buffers do not drive when  $V_{CCIO}$  is powered before  $V_{CCINT}$  or if the I/O pad voltage is higher than  $V_{CCIO}$ . This also applies for sudden voltage spikes during hot insertion. The  $V_{PAD}$  leakage current charges the 3.3-V tolerant circuit capacitance.

**Figure 4–2. Transistor-Level Diagram of MAX II Device I/O Buffers**



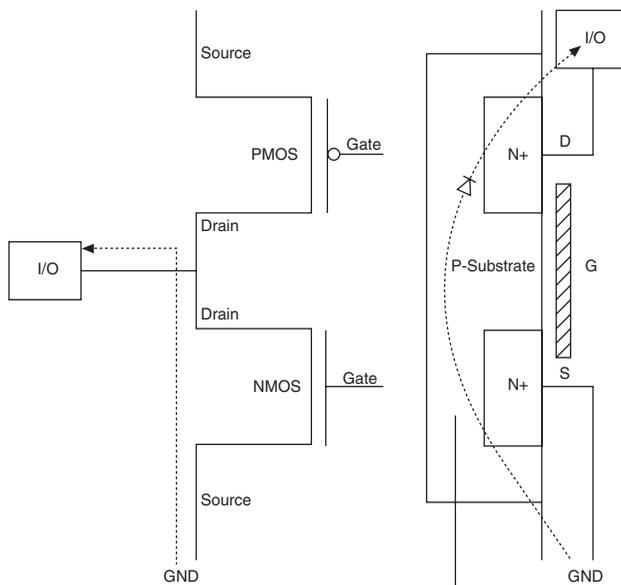
The CMOS output drivers in the I/O pins intrinsically provide electrostatic discharge (ESD) protection. There are two cases to consider for ESD voltage strikes: positive voltage zap and negative voltage zap.

A positive ESD voltage zap occurs when a positive voltage is present on an I/O pin due to an ESD charge event. This can cause the N+ (Drain)/P-Substrate junction of the N-channel drain to break down and the N+ (Drain)/P-Substrate/N+ (Source) intrinsic bipolar transistor turns on to discharge ESD current from I/O pin to GND. The dashed line (see [Figure 4–3](#)) shows the ESD current discharge path during a positive ESD zap.

**Figure 4–3. ESD Protection During Positive Voltage Zap**

When the I/O pin receives a negative ESD zap at the pin that is less than  $-0.7\text{ V}$  ( $0.7\text{ V}$  is the voltage drop across a diode), the intrinsic P-Substrate/N+ drain diode is forward biased. Hence, the discharge ESD current path is from GND to the I/O pin, as shown in [Figure 4–4](#).

Figure 4–4. ESD Protection During Negative Voltage Zap



## Power-On Reset Circuitry

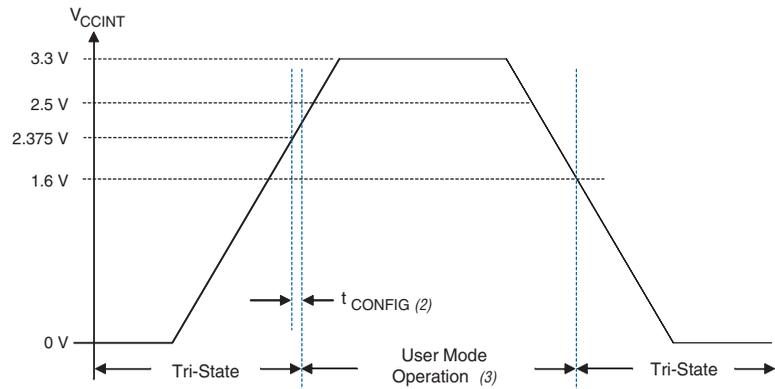
MAX II devices have POR circuits to  $V_{CCINT}$  and  $V_{CCIO}$  voltage levels during power-up. The POR circuit monitors these voltages triggering download from the non-volatile configuration flash memory (CFM) block to the SRAM logic maintaining tri-state of the I/O pins before and during this process. When the MAX II device enters user mode, the POR circuit releases the I/O pins to user functionality and continues to monitor the  $V_{CCINT}$  voltage level to detect a brown-out condition. If there is a  $V_{CCINT}$  voltage sag below the MAX II operational level during user mode, the POR circuit resets the device and re-triggers an SRAM download. The I/O bank  $V_{CCIO}$  levels are not monitored after initial power-up and transition into user mode functionality.

## Power-Up Characteristics

When power is applied to a MAX II device, the POR circuit monitors  $V_{CCINT}$  and begins SRAM download at a maximum voltage of 2.375 V, or 1.71 V for MAX II devices with a “G” ordering code. From this voltage reference, SRAM download and entry into user mode takes 100 to 300  $\mu$ s maximum depending on device density. Entry into user mode is gated by whether all  $V_{CCIO}$  banks are powered with sufficient operating voltage. If  $V_{CCINT}$  and  $V_{CCIO}$  are powered simultaneously, the device enters user mode within the 100 to 300  $\mu$ s specifications. If  $V_{CCIO}$  is powered more

than 100 to 300  $\mu\text{s}$  after  $V_{\text{CCINT}}$ , the device does not enter user mode until 2  $\mu\text{s}$  after all  $V_{\text{CCIO}}$  pins are powered. Figure 4-5 shows the device behavior during power-up.

**Figure 4-5. Power-Up Characteristics for MAX II Devices** Note (1)



**Notes to Figure 4-5:**

- (1) Figure 4-5 assumes all  $V_{\text{CCIO}}$  banks power simultaneously with the  $V_{\text{CCINT}}$  profile shown. If not,  $t_{\text{CONFIG}}$  stretches out until all  $V_{\text{CCIO}}$  banks are powered.
- (2) Voltages for  $t_{\text{CONFIG}}$  begin and user mode operation exit are maximum numbers.
- (3) The device is still operational for brown-out conditions down to 1.6 V, but does not meet specifications which are defined by the recommended operating condition ranges of  $V_{\text{CCINT}}$  and  $V_{\text{CCIO}}$ .



After SRAM configuration, all registers in the device are cleared and released into user function before I/O tri-states are released. To release clears after tri-states are released, use the `DEV_CLRn` pin option. To hold the tri-states beyond the power-up configuration time, use the `DEV_OE` pin option.



## Operating Conditions

Tables 5–1 through 5–12 provide information on absolute maximum ratings, recommended operating conditions, DC electrical characteristics, and other specifications for MAX<sup>®</sup> II devices.

### Absolute Maximum Ratings

Table 5–1 shows the absolute maximum ratings for the MAX II device family.

**Table 5–1. MAX II Device Absolute Maximum Ratings**    *Notes (1), (2)*

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
$V_{CCINT}$	Internal Supply voltage (3)	With respect to ground	–0.5	4.6	V
$V_{CCIO}$	I/O Supply Voltage		–0.5	4.6	V
$V_I$	DC input voltage		–0.5	4.6	V
$I_{OUT}$	DC output current, per pin		–25	25	mA
$T_{STG}$	Storage temperature	No bias	–65	150	°C
$T_{AMB}$	Ambient temperature	Under bias	–65	135	°C
$T_J$	Junction temperature	TQFP and BGA packages under bias		135	°C

**Notes to Table 5–1:**

- (1) See the *Operating Requirements for Altera Devices Data Sheet*.
- (2) Conditions beyond those listed in Table 5–1 may cause permanent damage to a device. Additionally, device operation at the absolute maximum ratings for extended periods of time may have adverse effects on the device.
- (3) Maximum  $V_{CCINT}$  for MAX II devices is 4.6 V. For MAX II devices with a “G” ordering code, it is 2.4 V.

## Recommended Operating Conditions

Table 5–2 shows the MAX II device family recommended operating conditions.

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
$V_{CCINT}$ (1)	3.3-V supply voltage for internal logic and ISP		3.00	3.60	V
	2.5-V supply voltage for internal logic and ISP		2.375	2.625	V
	1.8-V supply voltage for internal logic and ISP (MAX II devices with “G” ordering code)		1.71	1.89	V
$V_{CCIO}$ (1)	Supply voltage for I/O buffers, 3.3-V operation		3.00	3.60	V
	Supply voltage for I/O buffers, 2.5-V operation		2.375	2.625	V
	Supply voltage for I/O buffers, 1.8-V operation		1.71	1.89	V
	Supply voltage for I/O buffers, 1.5-V operation		1.425	1.575	V
$V_I$	Input voltage	(2), (3), (4)	–0.5	4.0	V
$V_O$	Output voltage		0	$V_{CCIO}$	V
$T_J$	Operating junction temperature	For commercial use	0	85	° C
		For industrial use	–40	100	° C
$t_R$	Input rise time			40	ns
$t_F$	Input fall time			40	ns
$t_{RSCHMITT}$	Schmitt trigger input rise time			(5)	

**Table 5–2. MAX II Device Recommended Operating Conditions (Part 2 of 2)**

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
$t_{\text{FSCHMITT}}$	Schmitt trigger input fall time			(5)	

**Notes to Table 5–2:**

- UFM programming via JTAG or logic array is not guaranteed outside the recommended operating conditions ( i.e., if brown-out occurs in the system during a potential write/program sequence to the UFM, users are recommended to read back UFM contents and verify against the intended write data).
- Minimum DC input is  $-0.5\text{ V}$ . During transitions, the inputs may undershoot to  $-2.0\text{ V}$  for input currents less than  $100\text{ mA}$  and periods shorter than  $20\text{ ns}$ .
- During transitions, the inputs may overshoot to the voltages shown in the following table based upon input duty cycle. The DC case is equivalent to 100% duty cycle. For more information on 5.0-V tolerance refer to the chapter on *Using MAX II Devices in Multi-Voltage Systems*.

$V_{\text{IN}}$	Max. Duty Cycle
4.0 V	100% (DC)
4.1	90%
4.2	50%
4.3	30%
4.4	17%
4.5	10%
- All pins, including clock, I/O, and JTAG pins, may be driven before  $V_{\text{CCINT}}$  and  $V_{\text{CCIO}}$  are powered.
- Contact Altera Applications for more information.

## Programming/Erase Specifications

Table 5–3 shows the MAX II device family programming/erase specifications.

**Table 5–3. MAX II Device Programming/Erase Specifications**

Parameter	Minimum	Typical	Maximum	Unit
Erase and reprogram cycles	100 (1)			Cycles

**Note to Table 5–3:**

- This specification applies to the user flash memory (UFM) and CFM blocks.

## DC Electrical Characteristics

Table 5–4 shows the MAX II device family DC electrical characteristics.

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
$I_I$	Input pin leakage current	$V_I = V_{CCIOmax}$ to 0 V (2)	–10		10	$\mu A$
$I_{OZ}$	Tri-stated I/O pin leakage current	$V_O = V_{CCIOmax}$ to 0 V (2)	–10		10	$\mu A$
$I_{CCSTANDBY}$	$V_{CCINT}$ supply current (standby) (3)	MAX II devices		12		$mA$
		MAX II Devices with a "G" ordering code		2		$mA$
$V_{SCHMITT}$	Hysteresis for Schmitt trigger input	$V_{CCIO} = 3.3$ V		460		$mV$
		$V_{CCIO} = 2.5$ V		170		$mV$
$I_{CCPower-up}$	$V_{CCINT}$ supply current during power-up (4)	MAX II devices		40		$mA$
		MAX II devices with a "G" ordering code		30		$mA$
$R_{PULLUP}$	Value of I/O pin pull-up resistor during power-up and in-system programming	$V_{CCIO} = 3.3$ V (5) $V_{CCIO} = 2.5$ V (5) $V_{CCIO} = 1.8$ V (5) $V_{CCIO} = 1.5$ V (5)	5	13	40	$k\Omega$
$C_{IO}$	Input capacitance for user I/O pin				8	$pF$
$C_{GCLK}$	Input capacitance for dual-purpose GCLK/user I/O pin				8	$pF$

**Note to Table 5–4:**

- (1) Typical values are for  $T_A = 25^\circ C$ ,  $V_{CCINT} = 3.3$  or  $2.5$  V, and  $V_{CCIO} = 1.5$  V,  $1.8$  V,  $2.5$  V, or  $3.3$  V.
- (2) This value is specified for normal device operation. The value may vary during power-up. This applies for all  $V_{CCIO}$  settings ( $3.3$ ,  $2.5$ ,  $1.8$ , and  $1.5$  V).
- (3)  $V_I =$  ground, no load, no toggling inputs.
- (4) This is average current during power-up. The typical peak current is no more than  $65$   $mA$  for MAX II devices. For MAX II devices with the "G" ordering code, the typical peak current is no more than  $55$   $mA$ .
- (5) Pin pull-up resistance values will lower if an external source drives the pin higher than  $V_{CCIO}$ .

## I/O Standard Specifications

Tables 5–5 through 5–10 show the MAX II device family I/O standard specifications.

**Table 5–5. 3.3-V LVTTTL Specifications**

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
$V_{CCIO}$	I/O supply voltage		3.0	3.6	V
$V_{IH}$	High-level input voltage		1.7	4.0	V
$V_{IL}$	Low-level input voltage		–0.5	0.8	V
$V_{OH}$	High-level output voltage	$I_{OH} = -8 \text{ mA}$ (1)	2.4		V
$V_{OL}$	Low-level output voltage	$I_{OL} = 8 \text{ mA}$ (1)		0.45	V

**Table 5–6. 3.3-V LVCMOS Specifications**

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
$V_{CCIO}$	I/O supply voltage		3.0	3.6	V
$V_{IH}$	High-level input voltage		1.7	4.0	V
$V_{IL}$	Low-level input voltage		–0.5	0.8	V
$V_{OH}$	High-level output voltage	$V_{CCIO} = 3.0$ , $I_{OH} = -0.1 \text{ mA}$	$V_{CCIO} - 0.2$		V
$V_{OL}$	Low-level output voltage	$V_{CCIO} = 3.0$ , $I_{OL} = 0.1 \text{ mA}$		0.2	V

**Table 5–7. 2.5-V I/O Specifications (Part 1 of 2)**

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
$V_{CCIO}$	I/O supply voltage		2.375	2.625	V
$V_{IH}$	High-level input voltage		1.7	4.0	V
$V_{IL}$	Low-level input voltage		–0.5	0.7	V

**Table 5–7. 2.5-V I/O Specifications (Part 2 of 2)**

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = –0.1 mA	2.1		V
		I <sub>OH</sub> = –1 mA	2.0		V
		I <sub>OH</sub> = –2 mA (1)	1.7		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 0.1 mA		0.2	V
		I <sub>OL</sub> = 1 mA		0.4	V
		I <sub>OL</sub> = 2 mA (1)		0.7	V

**Table 5–8. 1.8-V I/O Specifications**

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V <sub>CCIO</sub>	I/O supply voltage		1.71	1.89	V
V <sub>IH</sub>	High-level input voltage		0.65 × V <sub>CCIO</sub>	2.25	V
V <sub>IL</sub>	Low-level input voltage		–0.3	0.35 × V <sub>CCIO</sub>	V
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = –2 mA (1)	V <sub>CCIO</sub> – 0.45		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 2 mA (1)		0.45	V

**Table 5–9. 1.5-V I/O Specifications**

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V <sub>CCIO</sub>	I/O supply voltage		1.425	1.575	V
V <sub>IH</sub>	High-level input voltage		0.65 × V <sub>CCIO</sub>	V <sub>CCIO</sub> + 0.3	V
V <sub>IL</sub>	Low-level input voltage		–0.3	0.35 × V <sub>CCIO</sub>	V
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = –2 mA (1)	0.75 × V <sub>CCIO</sub>		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 2 mA (1)		0.25 × V <sub>CCIO</sub>	V

Notes to Tables 5–5 through 5–9:

(1) Drive strength is programmable according to values in Chapter 2, MAX II Architecture.

**Table 5–10. 3.3-V PCI Specifications**

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
$V_{CCIO}$	I/O supply voltage		3.0	3.3	3.6	V
$V_{IH}$	High-level input voltage		$0.5 \times V_{CCIO}$		$V_{CCIO} + 0.5$	V
$V_{IL}$	Low-level input voltage		-0.5		$0.3 \times V_{CCIO}$	V
$V_{OH}$	High-level output voltage	$I_{OH} = -500 \mu\text{A}$	$0.9 \times V_{CCIO}$			V
$V_{OL}$	Low-level output voltage	$I_{OL} = 1.5 \text{ mA}$			$0.1 \times V_{CCIO}$	V

### Bus Hold Specifications

Table 5–11 shows the MAX II device family bus hold specifications.

**Table 5–11. Bus Hold Specifications**

Parameter	Conditions	$V_{CCIO}$ Level								Unit
		1.5 V		1.8 V		2.5 V		3.3 V		
		Min	Max	Min	Max	Min	Max	Min	Max	
Low sustaining current	$V_{IN} > V_{IL}$ (maximum)	20		30		50		70		$\mu\text{A}$
High sustaining current	$V_{IN} < V_{IH}$ (minimum)	-20		-30		-50		-70		$\mu\text{A}$
Low overdrive current	$0 \text{ V} < V_{IN} < V_{CCIO}$		160		200		300		500	$\mu\text{A}$
High overdrive current	$0 \text{ V} < V_{IN} < V_{CCIO}$		-160		-200		-300		-500	$\mu\text{A}$

## Power-Up Timing

Table 5–12 shows the power-up timing characteristics for MAX II devices.

Symbol	Parameter	Device	Min	Typ	Max	Unit
$t_{\text{CONFIG}}$	The amount of time from when $V_{\text{CCINT}}$ reaches 2.375 V until the device enters user mode (2)	EPM240			100	$\mu\text{s}$
		EPM570			200	$\mu\text{s}$
		EPM1270			200	$\mu\text{s}$
		EPM2210			300	$\mu\text{s}$

### Notes to Table 5–12:

- (1) These numbers are preliminary.
- (2) For more information on POR trigger voltage, refer to the chapter on *Hot Socketing & Power-On Reset in MAX II Devices*.

## Power Consumption

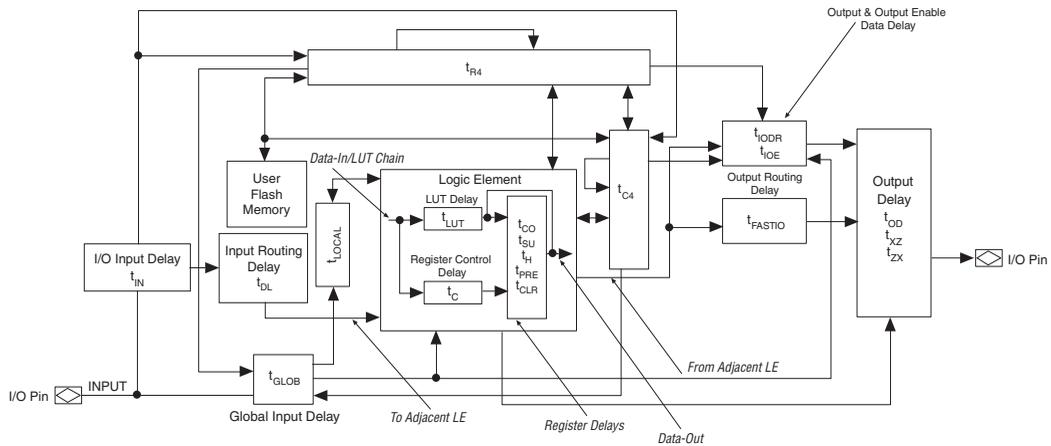
Designers can use the Altera® web power calculator to estimate the device power. See the chapter on *Understanding & Evaluating Power in MAX II Devices* for more information.

## Timing Model & Specifications

MAX II devices timing can be analyzed with the Altera Quartus II software, a variety of popular industry-standard EDA simulators and timing analyzers, or with the timing model shown in [Figure 5–1](#).

MAX II devices have predictable internal delays that enable the designer to determine the worst-case timing of any design. The software provides timing simulation, point-to-point delay prediction, and detailed timing analysis for device-wide performance evaluation.

Figure 5–1. MAX II Device Timing Model



The timing characteristics of any signal path can be derived from the timing model and parameters of a particular device. External timing parameters, which represent pin-to-pin timing delays, can be calculated as the sum of internal parameters. Refer to the chapter on *Understanding Timing in MAX II Devices* for more information.

This section describes and specifies the performance, internal, external, and UFM timing specifications. All specifications are representative of worst-case supply voltage and junction temperature conditions.

## Preliminary & Final Timing

Timing models can have either preliminary or final status. The Quartus® II software issues an informational message during the design compilation if the timing models are preliminary. Table 5–13 shows the status of the MAX II device timing models.

Preliminary status means the timing model is subject to change. Initially, timing numbers are created using simulation results, process data, and other known parameters. These tests are used to make the preliminary numbers as close to the actual timing parameters as possible.

Final timing numbers are based on actual device operation and testing. These numbers reflect the actual performance of the device under worst-case voltage and junction temperature conditions.

**Table 5–13. MAX II Device Timing Model Status**

Device	Preliminary	Final
EPM240	✓	
EPM570	✓	
EPM1270	✓	
EPM2210	✓	

### Performance

Table 5–14 shows the MAX II device performance for some common designs. All performance values were obtained with the Quartus II software compilation of megafunctions. These performance values are based on an EPM1270 device target.

**Table 5–14. MAX II Device Performance (Part 1 of 2)**

Resource Used	Design Size & Function	Mode	Resources Used		Performance			Unit
			LEs	UFM Blocks	-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	
LE	16-bit counter (1)	-	16	0	304	265.5	215.6	MHz
	64-bit counter (1)	-	66	0	154.2	118.8	96.5	MHz
	16-to-1 multiplexer	-	11	0	6.0	7.7	9.6	ns
	32-to-1 multiplexer	-	24	0	7.0	8.8	11.0	ns
	16-bit XOR function	-	5	0	5.2	6.7	8.2	ns
	16-bit decoder with single address line	-	5	0	5.2	6.7	8.2	ns

**Table 5–14. MAX II Device Performance (Part 2 of 2)**

Resource Used	Design Size & Function	Mode	Resources Used		Performance			Unit
			LEs	UFM Blocks	-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	
UFM	512 x 16	None	0	1	10.0	10.0	10.0	MHz
	512 x 16	SPI (2)	40	1	9.8	8.8	8.6	MHz
	512 x 8	Parallel (2)	73	1	(3)	(3)	(3)	MHz

**Notes to Table 5–14:**

- (1) This design is a binary loadable up counter.
- (2) This design is configured for read-only operation. Read and write ability increases the number of LEs used.
- (3) This design is asynchronous.

## Internal Timing Parameters

Internal timing parameters are specified on a speed grade basis independent of device density. Tables 5–15 through 5–22 describe the MAX II device internal timing microparameters for logic elements (LEs), input/output elements (IOEs), UFM structures, and MultiTrack™ interconnects.



For more explanations and descriptions on each internal timing microparameters symbol, refer to the chapter on *Understanding Timing in MAX II Devices*.

**Table 5–15. LE Internal Timing Microparameters (Part 1 of 2)**

Symbol	Parameter	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
t <sub>LUT</sub>	LE combinational LUT delay		582		756		931	ps
t <sub>CLR</sub>	LE register clear delay	243		315		388		ps
t <sub>PRE</sub>	LE register preset delay	243		315		388		ps
t <sub>SU</sub>	LE register setup time before clock	113		146		180		ps
t <sub>H</sub>	LE register hold time after clock	0		0		0		ps
t <sub>CO</sub>	LE register clock-to-output delay		243		315		388	ps

**Table 5–15. LE Internal Timing Microparameters (Part 2 of 2)**

Symbol	Parameter	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
$t_{CLKHL}$	Minimum clock high or low time	170		221		272		ps
$t_C$	Register control delay		875		1,137		1,400	ps

**Table 5–16. IOE Internal Timing Microparameters**

Symbol	Parameter	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
$t_{FASTIO}$	Data output delay from adjacent LE to I/O block		161		209		257	ps
$t_{IN}$	I/O input pad and buffer delay		727		945		1,163	ps
$t_{GLOB}$ (1)	I/O input pad and buffer delay use as global signal pin		1,413		1,836		2,259	ps
$t_{IOE}$	Internally generated output enable delay		424		552		679	ps
$t_{DL}$	Input routing delay		164		213		262	ps
$t_{OD}$ (2)	Output delay buffer and pad delay		1,454		1,890		2,326	ps
$t_{XZ}$ (3)	Output buffer disable delay		5,417		7,042		8,667	ps
$t_{ZX}$ (4)	Output buffer enable delay		1,404		1,825		2,246	ps

**Notes to Table 5–16:**

- (1) Delay numbers for  $t_{GLOB}$  differ for each device density and speed grade. The delay numbers shown in Table 5–16 are based on an EPM240 device target.
- (2) Refer to Table 5–28 and Table 5–29 for delay adders associated with different I/O Standards, drive strengths, and slew rates.
- (3) Refer to Table 5–19 and Table 5–20 for  $t_{XZ}$  delay adders associated with different I/O Standards, drive strengths, and slew rates.
- (4) Refer to Table 5–17 and Table 5–18 for  $t_{ZX}$  delay adders associated with different I/O Standards, drive strengths, and slew rates.

Tables 5–17 and 5–18 show the adder delays for  $t_{OD}$  and  $t_{ZX}$  microparameters when using an I/O standard other than 3.3-V LVTTTL with 16 mA drive strength.

**Table 5–17.  $t_{ZX}$  IOE Microparameter Adders for Fast Slew Rate**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
LVCMOS	4 mA		490		637		784	ps
	8 mA		0		0		0	ps
3.3-V LVTTTL	8 mA		490		637		784	ps
	16 mA		0		0		0	ps
2.5-V LVTTTL	7 mA		728		946		1,165	ps
	14 mA		146		190		234	ps
1.8-V LVTTTL	3 mA		2,176		2,829		3,482	ps
	6 mA		919		1,194		1,470	ps
1.5-V LVTTTL	2 mA		3,902		5,072		6,243	ps
	4 mA		1,682		2,186		2,691	ps

**Table 5–18.  $t_{ZX}$  IOE Microparameter Adders for Slow Slew Rate**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
LVCMOS	4 mA		9,176		8,755		8,334	ps
	8 mA		7,470		7,049		6,628	ps
3.3-V LVTTTL	8 mA		9,176		8,755		8,334	ps
	16 mA		7,470		7,049		6,628	ps
2.5-V LVTTTL	7 mA		13,366		12,945		12,524	ps
	14 mA		11,056		10,635		10,214	ps

**Table 5–19.  $t_{ZX}$  IOE Microparameter Adders for Fast Slew Rate (Part 1 of 2)**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
LVCMOS	4 mA		-367		-477		-587	ps
	8 mA		0		0		0	ps

**Table 5–19.  $t_{XZ}$  IOE Microparameter Adders for Fast Slew Rate (Part 2 of 2)**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
3.3-V LVTTTL	8 mA		–367		–477		–587	ps
	16 mA		0		0		0	ps
2.5-V LVTTTL	7 mA		–414		–539		–663	ps
	14 mA		–30		–39		–48	ps
1.8-V LVTTTL	3 mA		–588		–765		–941	ps
	6 mA		–33		–43		–53	ps
1.5-V LVTTTL	2 mA		–695		–904		–1,112	ps
	4 mA		95		123		152	ps

**Table 5–20.  $t_{XZ}$  IOE Microparameter Adders for Slow Slew Rate**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
LVCMOS	4 mA		–171		–1,796		–3,421	ps
	8 mA		135		–1,490		–3,115	ps
3.3-V LVTTTL	8 mA		–171		–1,796		–3,421	ps
	16 mA		135		–1,490		–3,115	ps
2.5-V LVTTTL	7 mA		–186		–1,811		–3,436	ps
	14 mA		164		–1,461		–3,086	ps

**Table 5–21. UFM Block Internal Timing Microparameters (Part 1 of 3)**

Symbol	Parameter	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
$t_{ASU}$	Address register shift signal setup to address register clock	20		20		20		ns
$t_{AH}$	Address register shift signal hold to address register clock	20		20		20		ns

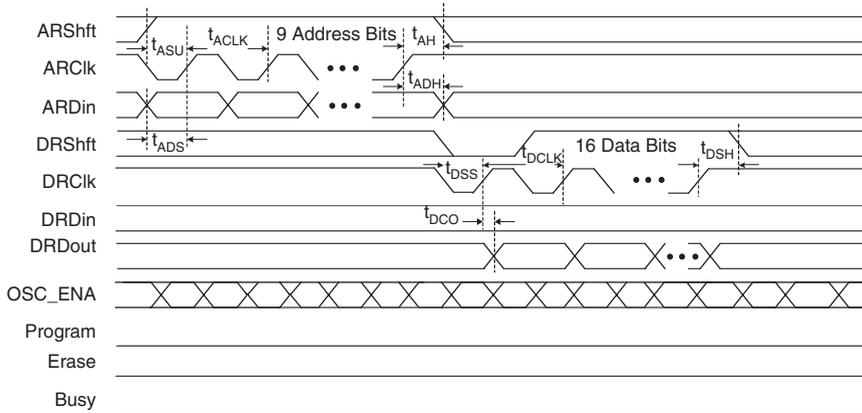
**Table 5–21. UFM Block Internal Timing Microparameters (Part 2 of 3)**

Symbol	Parameter	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
$t_{ADS}$	Address register data in setup to address register clock	20		20		20		ns
$t_{ADH}$	Address register data in hold from address register clock	20		20		20		ns
$t_{DSS}$	Data register shift signal setup to data register clock	40		40		40		ns
$t_{DSH}$	Data register shift signal hold from data register clock	20		20		20		ns
$t_{DDS}$	Data register data in setup to data register clock	20		20		20		ns
$t_{DDH}$	Data register data in hold from data register clock	20		20		20		ns
$t_{DP}$	Program signal to data clock hold time	0		0		0		ns
$t_{PB}$	Maximum delay between program rising edge to UFM busy signal rising edge		400		400		400	ns
$t_{BP}$	Minimum delay allowed from UFM busy signal going low to program signal going low	20		20		20		ns
$t_{PPMX}$	Maximum length of busy pulse during a program		75		75		75	$\mu$ s
$t_{AE}$	Minimum erase signal to address clock hold time	0		0		0		ns

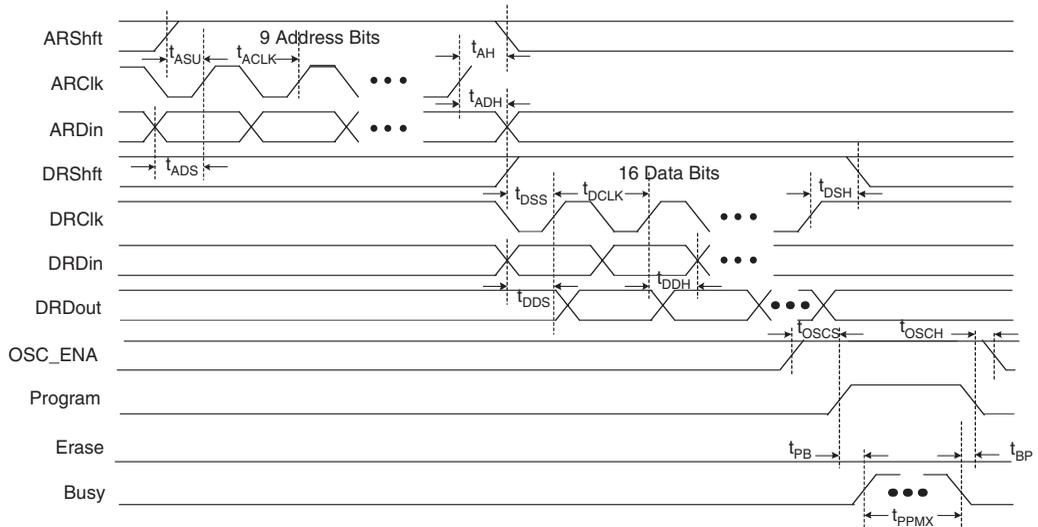
Symbol	Parameter	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
$t_{EB}$	Maximum delay between the erase rising edge to the UFM busy signal rising edge		400		400		400	ns
$t_{BE}$	Minimum delay allowed from the UFM busy signal going low to erase signal going low	20		20		20		ns
$t_{EPMX}$	Maximum length of busy pulse during an erase		36		36		36	ms
$t_{DCO}$	Delay from data register clock to data register output		50		50		50	ns
$t_{OE}$	Delay from data register clock to data register output	50		50		50		ns
$t_{RA}$	Maximum read access time		90		90		90	ns
$t_{OSCS}$	Maximum delay between the OSC_ENA rising edge to the erase/program signal rising edge	250		250		250		ns
$t_{OSCH}$	Minimum delay allowed from the erase/program signal going low to OSC_ENA signal going low	250		250		250		ns

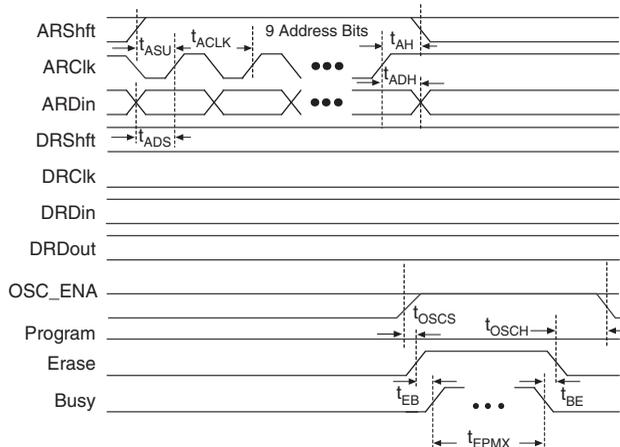
Figures 5–2 through 5–4 show the read, program, and erase waveforms for UFM block timing parameters shown in Table 5–22.

**Figure 5–2. UFM Read Waveforms**



**Figure 5–3. UFM Program Waveforms**



**Figure 5–4. UFM Erase Waveforms**

**Table 5–22. Routing Delay Internal Timing Microparameters**

Routing	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
$t_{C4}$		450		585		719	ps
$t_{R4}$		341		444		546	ps
$t_{LOCAL}$		342		445		548	ps

## External Timing Parameters

External timing parameters are specified by device density and speed grade. All external I/O timing parameters shown are for the 3.3-V LVTTTL I/O standard with the maximum drive strength and fast slew rate. For external I/O timing using standards other than LVTTTL or for different current strengths, use the I/O standard input and output delay adders in [Tables 5–27 through 5–29](#).

Table 5–23 shows the external I/O timing parameters for EPM240 devices.

**Table 5–23. EPM240 Global Clock External I/O Timing Parameters**

Symbol	Parameter	Condition	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
			Min	Max	Min	Max	Min	Max	
$t_{PD1}$	Worst case pin to pin delay through 1 look-up table (LUT)	10 pF		4.4		5.7		7.1	ns
$t_{PD2}$	Best case pin to pin delay through 1 LUT	10 pF		3.6		4.7		5.7	ns
$t_{SU}$	Global clock setup time		1.9		2.5		3.0		ns
$t_H$	Global clock hold time		0.0		0.0		0.0		ns
$t_{CO}$	Global clock to output delay	10 pF	2.0	4.2	2.0	5.4	2.0	6.7	ns
$t_{CH}$	Global clock high time		170		221		272		ps
$t_{CL}$	Global clock low time		170		221		272		ps
$t_{CNT}$	Minimum global clock period for 16-bit counter		3.3		3.8		4.6		ns
$f_{CNT}$	Maximum global clock frequency for 16-bit counter			304.0 (1)		265.5		215.6	MHz

**Note to Table 5–23:**

- (1) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

Table 5–24 shows the external I/O timing parameters for EPM570 devices.

**Table 5–24. EPM570 Global Clock External I/O Timing Parameters**

Symbol	Parameter	Condition	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
			Min	Max	Min	Max	Min	Max	
$t_{PD1}$	Worst case pin to pin delay through 1 look-up table (LUT)	10 pF		5.2		6.8		8.3	ns
$t_{PD2}$	Best case pin to pin delay through 1 LUT	10 pF		3.6		4.7		5.7	ns
$t_{SU}$	Global clock setup time		1.8		2.3		2.8		ns
$t_H$	Global clock hold time		0.0		0.0		0.0		ns
$t_{CO}$	Global clock to output delay	10 pF	2.0	4.3	2.0	5.6	2.0	6.9	ns
$t_{CH}$	Global clock high time		170		221		272		ps
$t_{CL}$	Global clock low time		170		221		272		ps
$t_{CNT}$	Minimum global clock period for 16-bit counter		3.3		3.8		4.6		ns
$f_{CNT}$	Maximum global clock frequency for 16-bit counter			304.0 (1)		265.5		215.6	MHz

**Note to Table 5–24:**

- (1) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

Table 5–25 shows the external I/O timing parameters for EPM1270 devices

Symbol	Parameter	Condition	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
			Min	Max	Min	Max	Min	Max	
$t_{PD1}$	Worst case pin to pin delay through 1 look-up table (LUT)	10 pF		6.3		8.2		10.1	ns
$t_{PD2}$	Best case pin to pin delay through 1 LUT	10 pF		3.6		4.7		5.7	ns
$t_{SU}$	Global clock setup time		1.7		2.2		2.7		ns
$t_H$	Global clock hold time		0.0		0.0		0.0		ns
$t_{CO}$	Global clock to output delay	10 pF	2.0	4.4	2.0	5.7	2.0	7.1	ns
$t_{CH}$	Global clock high time		170		221		272		ps
$t_{CL}$	Global clock low time		170		221		272		ps
$t_{CNT}$	Minimum global clock period for 16-bit counter		3.3		3.8		4.6		ns
$f_{CNT}$	Maximum global clock frequency for 16-bit counter			304.0 (1)		265.5		215.6	MHz

**Note to Table 5–25:**

- (1) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

Table 5–26 shows the external I/O timing parameters for EPM2210 devices.

**Table 5–26. EPM2210 Global Clock External I/O Timing Parameters**

Symbol	Parameter	Condition	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
			Min	Max	Min	Max	Min	Max	
$t_{PD1}$	Worst case pin to pin delay through 1 look-up table (LUT)	10 pF		6.9		8.9		11.0	ns
$t_{PD2}$	Best case pin to pin delay through 1 LUT	10 pF		3.6		4.7		5.7	ns
$t_{SU}$	Global clock setup time		1.6		2.1		2.6		ns
$t_H$	Global clock hold time		0.0		0.0		0.0		ns
$t_{CO}$	Global clock to output delay	10 pF	2.0	4.5	2.0	5.8	2.0	7.2	ns
$t_{CH}$	Global clock high time		170		221		272		ps
$t_{CL}$	Global clock low time		170		221		272		ps
$t_{CNT}$	Minimum global clock period for 16-bit counter		3.3		3.8		4.6		ns
$f_{CNT}$	Maximum global clock frequency for 16-bit counter			304.0 (1)		265.5		215.6	MHz

**Note to Table 5–26:**

- (1) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

## External Timing I/O Delay Adders

I/O delay timing parameters for I/O standard input and output adders and input delays are specified by speed grade independent of device density.

Tables 5–27 through 5–29 show the adder delays associated with I/O pins for all packages. If an I/O standard is selected other than LVTTTL with a unit value of 16 mA and a fast slew rate, add the selected input delay adder to the external  $t_{SU}$  timing parameters shown in Tables 5–23 through 5–26. Add the output delay adder to the external  $t_{CO}$  and  $t_{PD}$  shown in Tables 5–23 through 5–26.

**Table 5–27. External Timing Input Delay Adders**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
3.3-V LVTTTL	Without Schmitt Trigger		0		0		0	ps
	With Schmitt Trigger		308		400		493	ps
3.3-V LVCMOS	Without Schmitt Trigger		0		0		0	ps
	With Schmitt Trigger		308		400		493	ps
2.5-V LVTTTL	Without Schmitt Trigger		16		20		25	ps
	With Schmitt Trigger		303		394		485	ps
1.8-V LVTTTL	Without Schmitt Trigger		273		355		437	ps
1.5-V LVTTTL	Without Schmitt Trigger		670		871		1,072	ps
3.3-V PCI	Without Schmitt Trigger		6		7		9	ps

**Table 5–28. External Timing Output Delay &  $t_{OD}$  Adders for Fast Slew Rate (Part 1 of 2)**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
3.3-V LVTTTL	16 mA		0		0		0	ps
	8 mA		404		525		646	ps

**Table 5–28. External Timing Output Delay &  $t_{OD}$  Adders for Fast Slew Rate (Part 2 of 2)**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
3.3-V LVCMOS	8 mA		0		0		0	ps
	4 mA		404		525		646	ps
2.5-V LVTTTL	14 mA		209		271		334	ps
	7 mA		681		885		1,090	ps
1.8-V LVTTTL	6 mA		878		1,141		1,405	ps
	3 mA		1,656		2,153		2,650	ps
1.5-V LVTTTL	4 mA		1641		2,133		2,626	ps
	2 mA		2,709		3,521		4,334	ps

**Table 5–29. External Timing Output Delay &  $t_{OD}$  Adders for Slow Slew Rate**

Standard		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
3.3-V LVTTTL	16 mA		7,085		6,649		6,213	ps
	8 mA		8,437		8,001		7,565	ps
3.3-V LVCMOS	8 mA		7,085		6,649		6,213	ps
	4 mA		8,437		8,001		7,565	ps
2.5-V LVTTTL	14 mA		10,356		9,920		9,484	ps
	7 mA		12,066		11,630		11,194	ps

## Maximum Input & Output Clock Rates

Tables 5–30 and 5–31 show the maximum input and output clock rates for standard I/O pins in MAX II devices.

**Table 5–30. MAX II Maximum Input Clock Rate for I/O (Part 1 of 2)**

Standard	-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
3.3-V LVTTTL	304	304	304	MHz
3.3-V LVCMOS	304	304	304	MHz
2.5-V LVTTTL	220	220	220	MHz
2.5-V LVCMOS	220	220	220	MHz
1.8-V LVTTTL	200	200	200	MHz
1.8-V LVCMOS	200	200	200	MHz

**Table 5–30. MAX II Maximum Input Clock Rate for I/O (Part 2 of 2)**

Standard	-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
1.5-V LVCMOS	150	150	150	MHz
PCI	33.3	33.3	33.3	MHz
3.3-V SCHMITT	250	250	250	MHz
2.5-V SCHMITT	188	188	188	MHz

**Table 5–31. MAX II Maximum Output Clock Rate for I/O**

Standard	-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	Unit
3.3-V LVTTTL	304	304	304	MHz
3.3-V LVCMOS	304	304	304	MHz
2.5-V LVTTTL	220	220	220	MHz
2.5-V LVCMOS	220	220	220	MHz
1.8-V LVTTTL	200	200	200	MHz
1.8-V LVCMOS	200	200	200	MHz
1.5-V LVCMOS	150	150	150	MHz
PCI	33.3	33.3	33.3	MHz

## JTAG Timing Specifications

Figure 5–5 shows the timing waveforms for the JTAG signals.

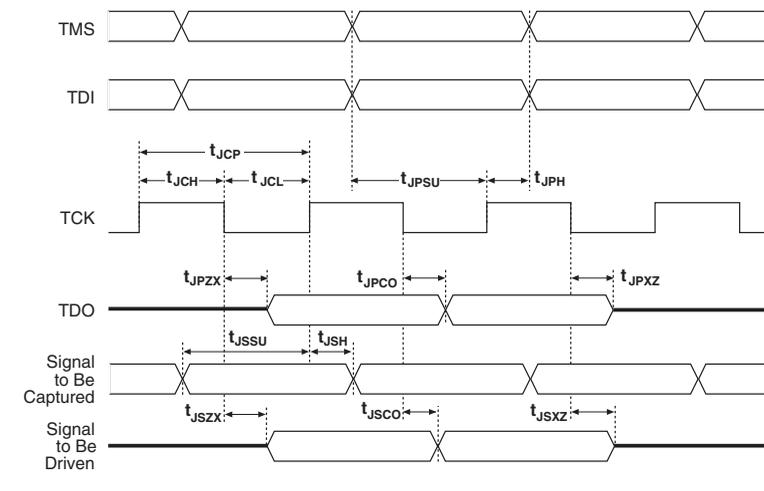
**Figure 5–5. MAX II JTAG Timing Waveforms**


Table 5–32 shows the JTAG Timing parameters and values for MAX II devices.

Symbol	Parameter	Min	Max	Unit
$t_{JCP}$	TCK clock period	40		ns
$t_{JCH}$	TCK clock high time	20		ns
$t_{JCL}$	TCK clock low time	20		ns
$t_{JPSU}$	JTAG port setup time (1)	8		ns
$t_{JPH}$	JTAG port hold time	10		ns
$t_{JPCO}$	JTAG port clock to output (1)		10	ns
$t_{JPZX}$	JTAG port high impedance to valid output		10	ns
$t_{JPXZ}$	JTAG port valid output to high impedance		10	ns
$t_{JSSU}$	Capture register setup time	8		ns
$t_{JSH}$	Capture register hold time	10		ns

**Table 5–32. MAX II JTAG Timing Parameters (Part 2 of 2)**

Symbol	Parameter	Min	Max	Unit
$t_{USCO}$	Update register clock to output		25	ns
$t_{USZX}$	Update register high impedance to valid output		25	ns
$t_{USXZ}$	Update register valid output to high impedance		25	ns

**Note to Table 5–32:**

- (1) This specification is shown for 3.3-V LVTTTL/LVCMOS and 2.5-V LVTTTL/LVCMOS operation of the JTAG pins. For 1.8-V LVTTTL/LVCMOS and 1.5-V LVCMOS, the JTAG port clock setup time is 6 ns and port clock to output time is 12 ns.



## Software

MAX<sup>®</sup> II devices are supported by the Altera<sup>®</sup> Quartus<sup>®</sup> II design software with new, optional MAX+PLUS<sup>®</sup> II look and feel, which provides HDL and schematic design entry, compilation and logic synthesis, full simulation and advanced timing analysis, and device programming. See the *Design Software Selector Guide* for more details on the Quartus II software features.

The Quartus II software supports the Windows XP/2000/NT, Sun Solaris, Linux Red Hat v8.0, and HP-UX operating systems. It also supports seamless integration with industry-leading EDA tools through the NativeLink<sup>®</sup> interface.

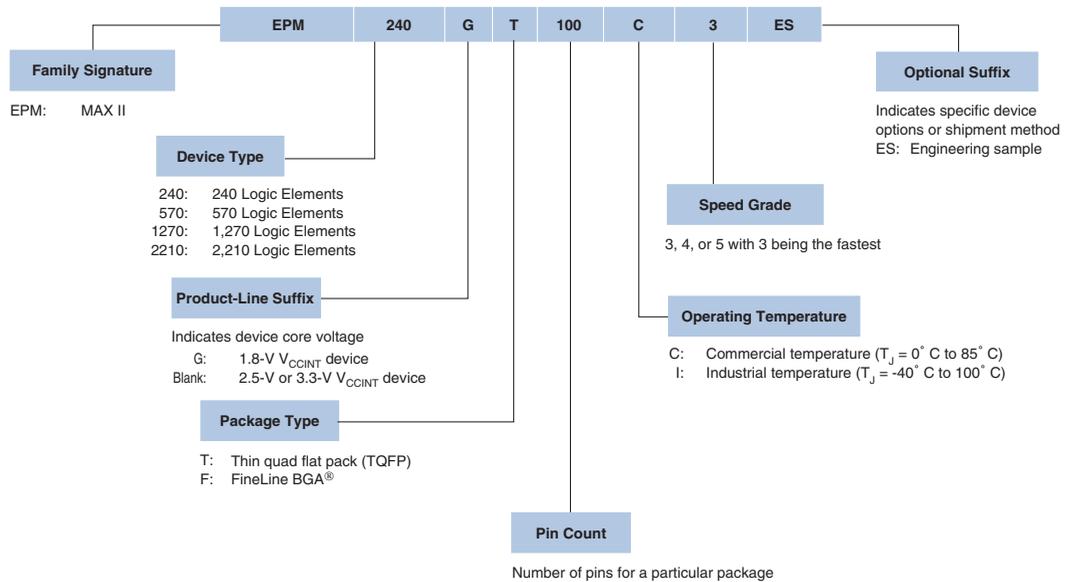
## Device Pin-Outs

Printed device pin-outs for MAX II devices will be released on the Altera web site ([www.altera.com](http://www.altera.com)) and in the MAX II Device Handbook when they are available.

## Ordering Information

[Figure 6-1](#) describes the ordering codes for MAX II devices. For more information on a specific package, refer to the chapter on *Package Information*.

**Figure 6–1. MAX II Device Packaging Ordering Information**



## Dual Marking

On MAX II devices, packages display a dual marking for the -4 commercial and -5 industrial ordering codes. For example, both EPM570GT100C4 and EPM570GT100I5 ordering codes are marked on the same package.